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A GRAPHICAL METHOD FOR CALCULATING GROUND REFLECTION COEFFICIENTS

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In problems involving radiation from and reception on radio antennas it is usually necessary to obtain the plane-wave reflection coefficients of the earth for vertically and horizontally polarized waves. These are designated as R_V and R_H , respectively.

Since the earth is an imperfectly conducting dielectric the effective dielectric constant is a complex number. Hence the refractive index is also a complex number, the imaginary part of which is a measure of the absorption of the wave in the ground, and of the phase change of the wave upon reflection from the ground. The evaluation of R_V and R_H is thus a rather tedious process, involving transformations of complex numbers. Under certain simplifying assumptions, however, which are justified to the degree of accuracy with which the ground constants are known, a graphical representation of R_V and R_H can be made, which assists in their calculation.

The reflection coefficients of the earth, for a plane wave whose direction of propagation makes a vertical angle θ with the ground, are, for the vector directions of Fig. 1:

$$R_V = \frac{u \sqrt{1-u^2 \cos^2 \theta} - \sin \theta}{u \sqrt{1-u^2 \cos^2 \theta} + \sin \theta} \quad (\text{for an electric vector in the plane of propagation})$$

and

$$R_H = \frac{u \sin \theta - \sqrt{1-u^2 \cos^2 \theta}}{u \sin \theta + \sqrt{1-u^2 \cos^2 \theta}} \quad (\text{for an electric vector perpendicular to the plane of propagation}).$$

For ground reflection these refer to vertically and horizontally polarized waves, respectively. In the formulas:

$$u^2 = \frac{1}{\epsilon - jx} = (\alpha + j\beta)^2$$

= specific inductive capacity of earth

$$x = \frac{18000 \gamma}{f}$$

γ = conductivity of earth in mhos per meter.

(over)

For most types of soil u^2 is not greater than 0.2; less than 5% error is introduced into $1 \pm R_V$ and $1 \pm R_H$ by neglecting $u^2 \cos^2 \theta$ in comparison with 1. (A considerably greater error may occur in R_V near Brewster's angle, however, for high frequencies and low conductivities).

Thus, to a good approximation, we may consider:

$$R_V = \frac{-u - \sin \theta}{u + \sin \theta}, \quad 1 + R_V = \frac{2u}{u + \sin \theta}, \quad 1 - R_V = \frac{2}{u \csc \theta + 1}$$

$$R_H = \frac{u \sin \theta - 1}{u \sin \theta + 1}, \quad 1 + R_H = \frac{2u}{u + \csc \theta}, \quad 1 - R_H = \frac{2}{u \sin \theta + 1}$$

The graphical construction shown in Fig. 2 illustrates the relations between numerators and denominators of R_V and R_H . Let OA be unit, AF and AF' be equal to $\frac{1}{|u|}$ and ϕ = the phase angle of $\frac{1}{u}$. Lay off the length $n = \sin \phi$ perpendicular to FAF'; then $m = \cos \phi$. Draw FD and F'D' parallel to OA and equal to unity.

If $AB = AB' = \frac{1}{|u|} \sin \theta$, then $\frac{OB}{OB'} = |R_V|$ and angle $BOB' =$ phase of R_V .

If $DC = D'C' = \sin \theta$, then $\frac{OC}{OC'} = |R_H|$ and angle $COC' =$ phase of R_H .

$$\text{Now } OB^2 = (AB - m)^2 + n^2 \quad \text{and } OB'^2 = (AB + m)^2 + n^2$$

To find the minimum value of $\frac{OB}{OB'}$ set

$$\frac{\partial}{\partial AB} \log \frac{OB^2}{OB'^2} = 0 \approx \frac{2(AB-m)}{(AB-m)^2 + n^2} - \frac{2(AB+m)}{(AB+m)^2 + n^2} = \frac{2(AB-m)}{AB^2 - 2AB + 1} - \frac{2(AB+m)}{AB^2 + 2AB + 1}$$

This will be the case if:

$$2(AB-m)(AB^2 + 2AB + 1) = 2(AB+m)(AB^2 - 2AB + 1)$$

i.e., if $AB = 1$ and therefore = OA.

Now when $AB = AB' = OA$, angle $BOB' = 90^\circ$ and $|u| = \sin \theta$

Therefore $R_V =$ minimum and its phase = 90° when $|u| = \sin \theta$

The minimum value of R_V is therefore given by:

$$R_{\min} = \frac{\alpha - \sqrt{\alpha^2 + \beta^2} + j\beta}{\alpha + \sqrt{\alpha^2 + \beta^2} + j\beta} = \frac{\alpha^2 - (\alpha^2 + \beta^2) + 2j\beta\sqrt{\alpha^2 + \beta^2} + \beta^2}{\alpha^2 + (\alpha^2 + \beta^2) + 2\alpha\sqrt{\alpha^2 + \beta^2} + \beta^2}$$

$$= \frac{j\beta}{\alpha + \sqrt{\alpha^2 + \beta^2}} = \frac{\beta}{\alpha + \sqrt{\alpha^2 + \beta^2}} \quad \angle 90^\circ$$

Now ϕ , the phase angle of $\frac{1}{u}$, $\dots \tan^{-1} \frac{\beta}{\alpha} = -\sin^{-1} \frac{\beta}{\sqrt{\alpha^2 + \beta^2}} = \cos^{-1} \frac{\alpha}{\sqrt{\alpha^2 + \beta^2}}$

$$|\tan \frac{1}{2}\phi| = \frac{\frac{\beta}{\sqrt{\alpha^2 + \beta^2}}}{1 + \frac{\alpha}{\sqrt{\alpha^2 + \beta^2}}} = |R_{\min}|$$

Further,

$$1 - R_V = \frac{2 \sin \theta}{u + \sin \theta} = \frac{2 \sin \theta}{\sqrt{(\alpha + \sin \theta)^2 + \beta^2}} \angle \phi_1 \quad \text{where } \phi_1 = -\tan^{-1} \frac{\beta}{\alpha \sin \theta}$$

and, if R_{V0} = value of R_V for $\theta = 0$

$$1 - R_{V0} = 0 \angle \phi_0 \quad \text{where } \phi_0 = -\tan^{-1} \frac{\beta}{\alpha}$$

Therefore

$$\begin{aligned} \sin (\phi_1 - \phi_0) &= \frac{\beta (\alpha + \sin \theta) - \alpha \beta}{\sqrt{(\alpha + \sin \theta)^2 + \beta^2} \sqrt{\alpha^2 + \beta^2}} = \frac{\sin \theta}{\sqrt{(\alpha + \sin \theta)^2 + \beta^2}} \cdot \frac{\beta}{\sqrt{\alpha^2 + \beta^2}} \\ &= \frac{\beta}{2\sqrt{\alpha^2 + \beta^2}} |1 - R_V| \end{aligned}$$

and thus $1 - R_V$ lies on a circle in the complex plane, and therefore R_V itself lies on a circle.

Similarly,

$$1 - R_H = \frac{2}{u \sin \theta + 1} = \frac{2}{\sqrt{(\alpha \sin \theta + 1)^2 + \beta^2 \sin^2 \theta}} \phi_2 \text{ where } \phi_2 = -\tan^{-1} \frac{\beta \sin \theta}{\alpha \sin \theta + 1}$$

and so it may be shown that

$$\sin(\phi_2 - \phi_0) = \frac{\beta}{\sqrt{\alpha^2 + \beta^2}} |1 - R_H|$$

so that $1 - R_H$, and hence also R_H , lies on the same circle as R_V .

This leads at once to the obvious graphical construction for R_V and R_H shown in Fig. 3. On the perpendicular bisector of a line of length 2 lay off the length $R_{\min} = \tan \frac{1}{2} \theta$ upward. From the top of this length drop down a

distance $\rho = \frac{1 + R_{\min}^2}{2R_{\min}}$ to locate the center of a circle of radius ρ , and

draw the arc of this circle of which the base line of length 2 is the chord.

To find R_V lay off the angle $\psi_1 = 2\beta_1 = 2 \tan^{-1} \frac{\beta}{\alpha + \sin \theta}$ and draw R_V from the center of the base line to the circle as shown. To find R_H

lay off the angle $\psi_2 = 2\phi_2 = 2 \tan^{-1} \frac{\beta \sin \theta}{\alpha \sin \theta + 1}$ and draw R_H in the

above manner. This comes about from the fact that the arc ψ is subtended by the phase angle of $1 - R$ and is therefore twice that angle. The nomograms of Figs. 4, 5, & 6 are attached for ready calculation of ψ_1 and ψ_2 .

The quantity $1 - R$ is found by drawing a line from A to the point where R intersects the circle, and the phase of $1 - R$ is minus the angle measured clockwise from line AB to the line $1 - R$. The quantity $1 + R$ is found by drawing a line from B to the point where R intersects the circle, and the phase angle is that measured counter-clockwise from line AB to the line $1 + R$.

It was mentioned above that the greatest error is likely to occur in the values of R_V near the Brewster angle θ , i.e., the value of θ for which R_V is a minimum. This error is largely in the angle θ_B itself rather than in the variation of R_V with θ . A better value of θ may be calculated from the formula

$$\theta_B = \sin^{-1} u \left[1 - \frac{u^2}{2(1 + |u|^2)} \right]$$

if more precise values of R_V near Brewster's angle are required.

Although this method of determining R_H and R_V is quantitatively valid only for materials with dielectric constants and conductivities, the representation of R_H and R_V as vectors lying on a single curve, as a Fig. 3, is generally useful, for example in optics, in showing how the reflection coefficients vary with angle. In the case of a perfect reflector, both the R_H and R_V vectors lie at the left-hand end of the curve (point B) for all values of θ except that R_V is indeterminate at $\theta = 0$.

In the case of a perfect dielectric ($x = 0$)

$$u = \sqrt{\frac{1}{\epsilon}} \quad , \quad \alpha = \sqrt{\frac{1}{\epsilon}} \quad , \quad \beta = 0 \quad , \quad R_{\min} = \lim_{\epsilon \rightarrow 0} \frac{\epsilon}{2\alpha} \quad , \quad \rho = \lim_{\epsilon \rightarrow 0} \frac{\alpha}{\epsilon}$$

and $1 = \lim_{\epsilon \rightarrow 0} \frac{2\epsilon}{\alpha + \sin \theta}$ $2 = \lim_{\epsilon \rightarrow 0} \frac{2\epsilon}{\alpha + \cos \theta}$ and the curve of

Fig. 3 is a straight line between B and A and R_V and R_H have phases of either 0° or 180° .

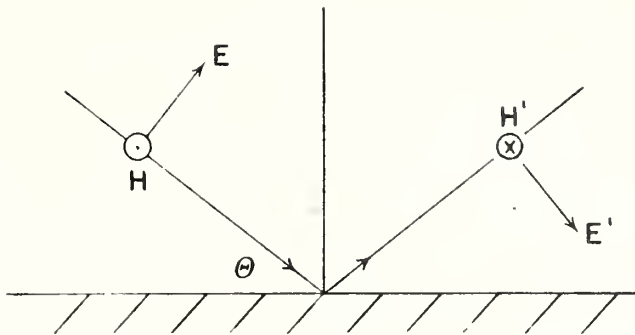
Thus $\rho\psi_1 = \frac{2\alpha}{\alpha + \sin \theta}$, and $\rho\psi_2 = \frac{2\alpha}{\alpha + \csc \theta}$, represent the distances

from B along the straight line of Fig. 2 to the ends of the vectors R_V

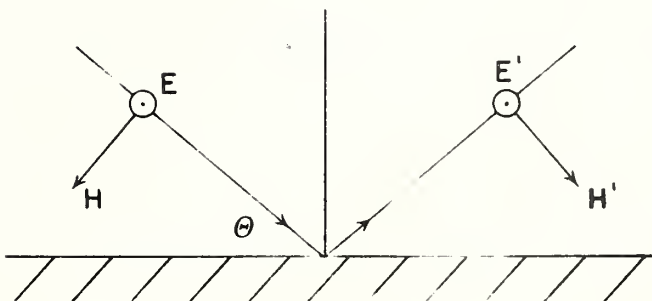
and R_H respectively. At $\theta = 90^\circ$ $\rho\psi_1 = \rho\psi_2 = \frac{2\alpha}{1 + \alpha}$ and $R_V, R_H = \frac{1 - \alpha}{1 + \alpha}$

$= \frac{\sqrt{\epsilon} - 1}{\sqrt{\epsilon} + 1}$ as is the usual formula for normal incidence. At Brewster's

angle $\sin \theta = \alpha$ and $\rho\psi_1 = 1$ giving $R_V = 0$. For θ less than Brewster's angle $\rho\psi_1 > 1$ and so R_V lies to the right and has a phase 0° ; for θ greater than Brewster's angle lies to the left, and for any angle R_H lies to the left, and so has a phase of 180° .



A. FOR VERTICALLY POLARIZED ELECTRIC FIELD.



B. FOR HORIZONTALLY POLARIZED ELECTRIC FIELD.

Fig. 1. DIRECTION CONVENTIONS FOR ELECTRIC AND MAGNETIC VECTORS.

⊙ = DIRECTED OUT OF PAPER.

⊗ = DIRECTED INTO PAPER.

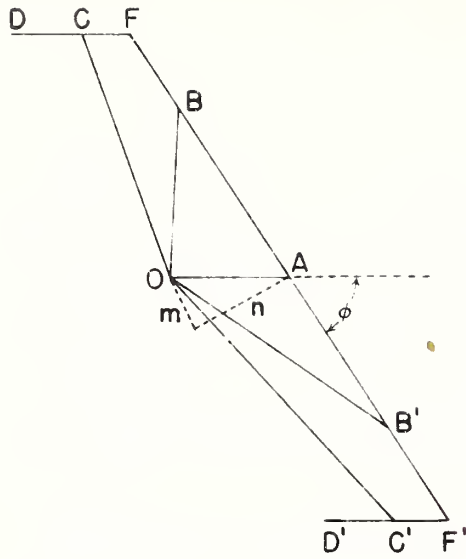


Fig. 2. GRAPHICAL REPRESENTATION OF NUMERATORS AND DENOMINATORS OF R_V AND R_H .

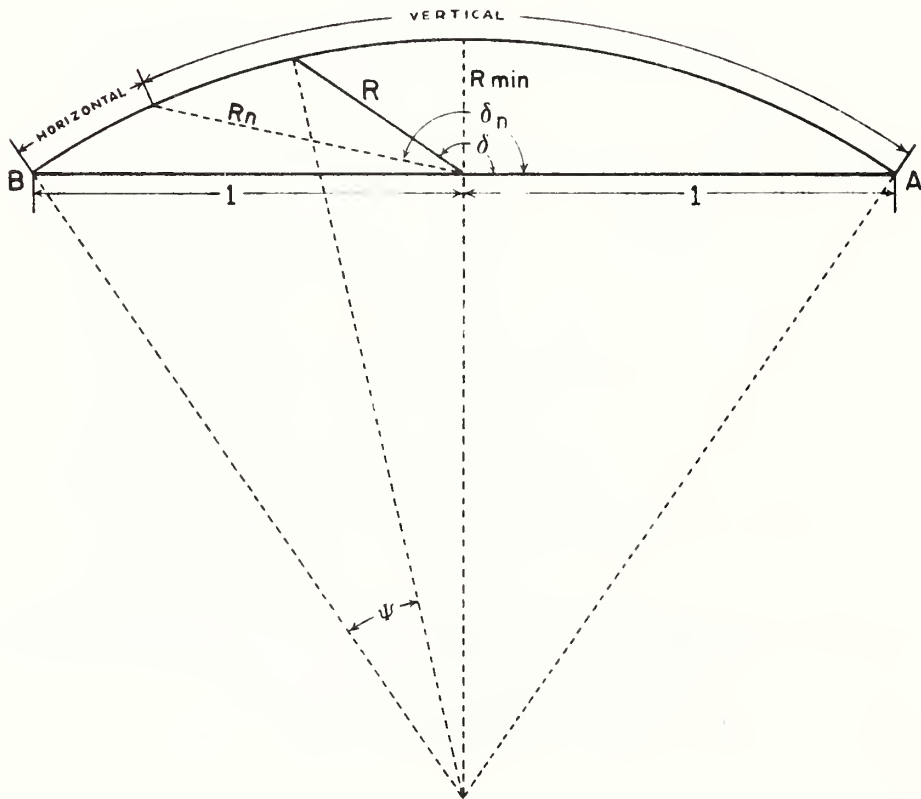


Fig. 3. GRAPHICAL CONSTRUCTION FOR R_V AND R_H .

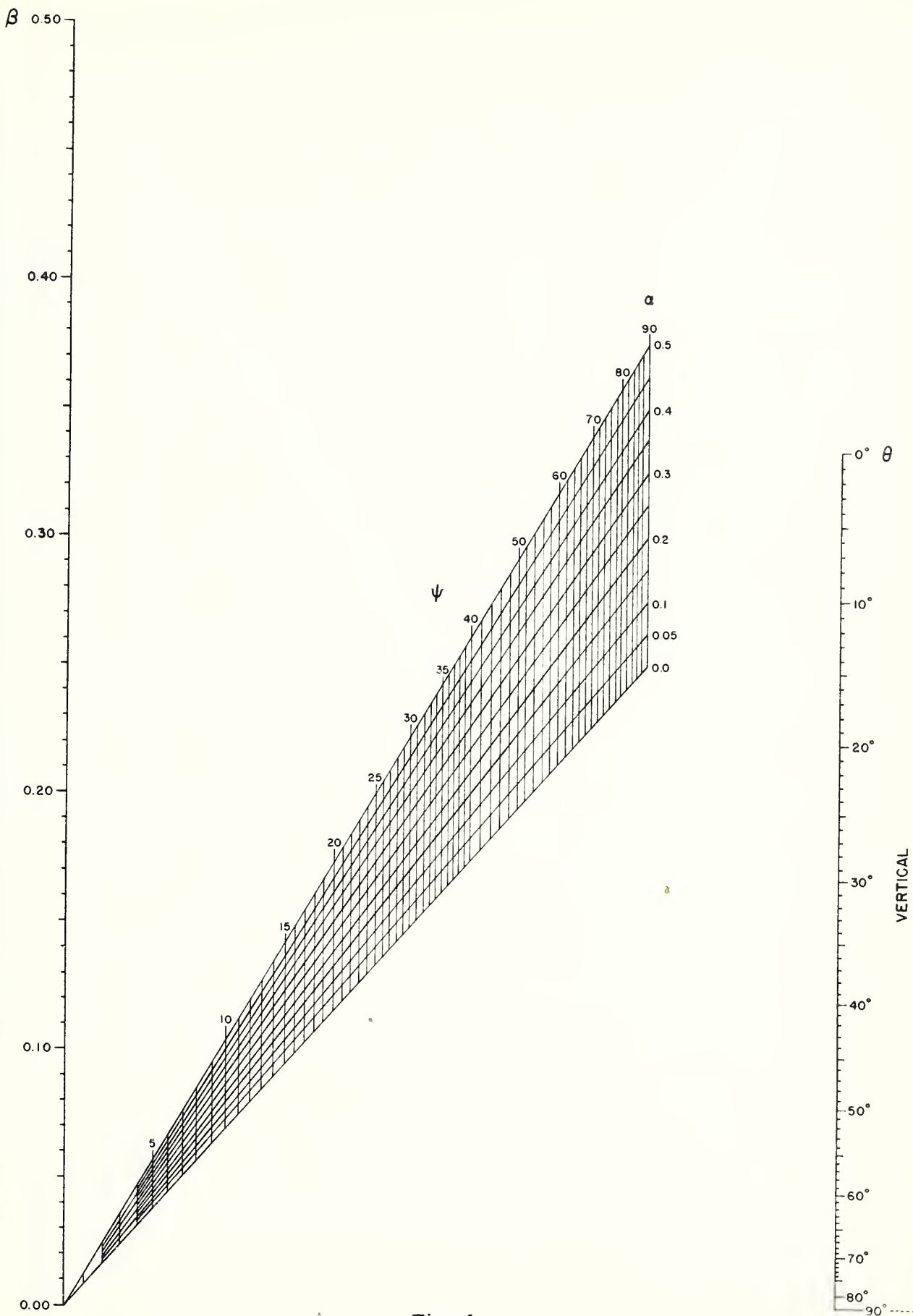


Fig. 4.

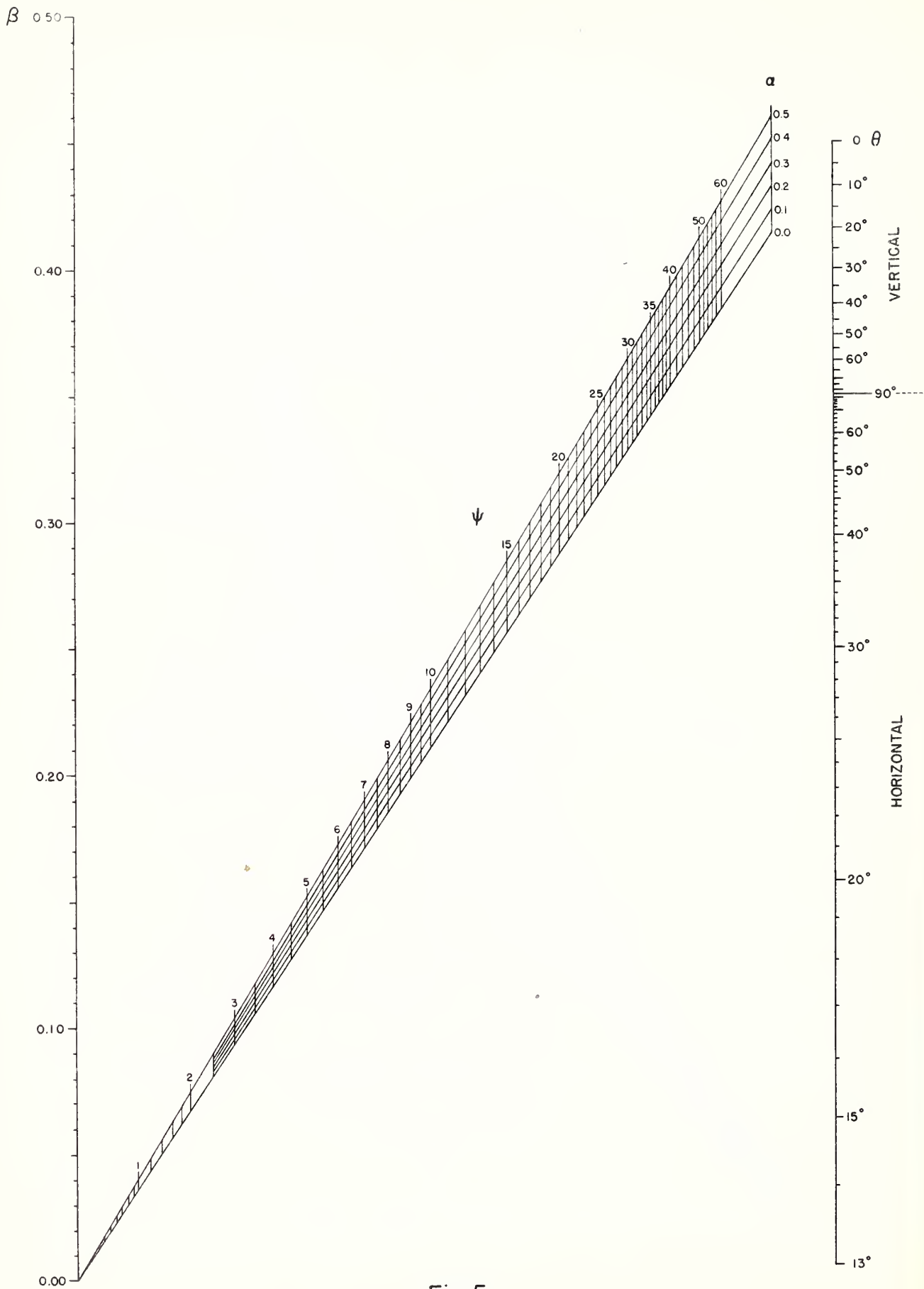


Fig. 5.

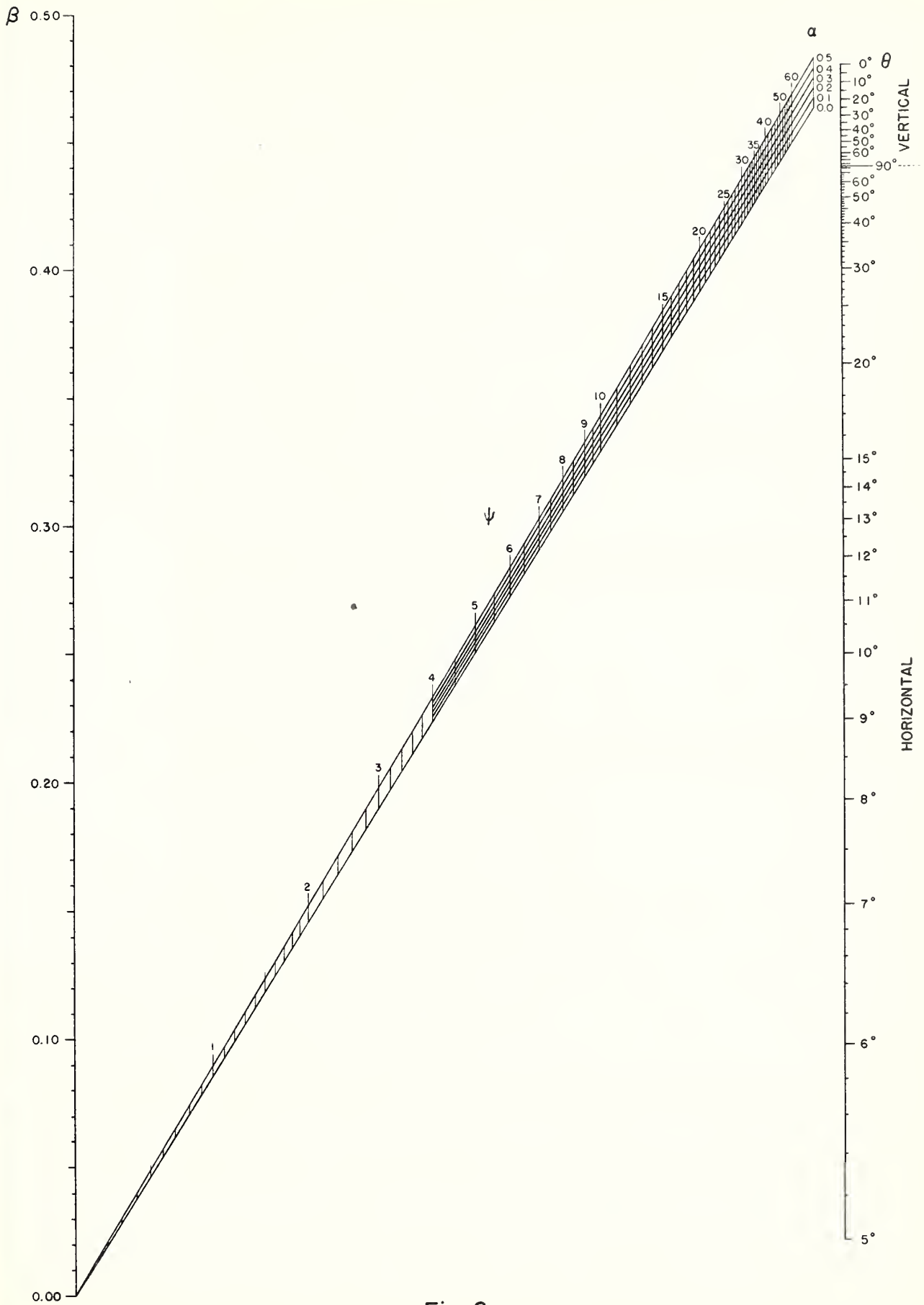


Fig. 6.

