
**EVALUATION OF DIELECTRIC PROPERTIES
OF POLYMER THIN-FILM MATERIALS
FOR APPLICATION IN EMBEDDED
CAPACITANCE**

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EXECUTIVE BRIEF

Evaluation of Dielectric Properties of Polymer Thin-Film Materials for Application in Embedded Capacitance

The relative dielectric constant of the embedded capacitance materials was measured in the frequency range of 100 Hz to 5 GHz. The testing included evaluation of the capacitance density, leakage current, and the effect of environmental stress on the capacitance. The objective of this project was to develop and evaluate a practical test method suitable for dielectric permittivity of high-k polymer composite films that covers a broad frequency range including the microwave. A preliminary set of test pattern, specification and corresponding testing procedure have been designed for dielectric characterization of the embedded capacitance materials. We have used them to compare the dielectric constant of several experimental high-k films recently developed by the industry. The low frequency test vehicle consists of lumped elements for the permittivity in z and x-y directions. The high-frequency test vehicle was designed as a two-layer circuitry with a number of microstrip resonators, transmission lines and coaxial terminations. The testing procedure has been examined on films ranging from 40 μm to 100 μm thick with relative dielectric constant ranging from 4 to 40. We found that the upper frequency limit of the measurements decreases with increasing value of the dielectric constant. The limit is about 18 GHz for films with the relative dielectric constant of 4, and decreases to about 5 GHz for films with the relative dielectric constant of 50.

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Chapter 1

INTRODUCTION

Polymer-based high dielectric constant films (high-K) can be used to construct embedded, discrete RLC circuits and decoupling power planes for use in wireless communication and high-speed electronics. The permittivity of the prospective materials should be high and determined with a high degree of confidence for operation at microwave frequencies. In order to develop and successfully commercialize such materials, the industry needs a suitable test method to measure dielectric properties and to assess the functional and reliability performance of these materials in planar, thin film configuration. National Center for Manufacturing Science (NCMS) has setup the Embedded Decoupling Capacitance Consortium to study this problem. The objectives for National Institute of Standards and Technology (NIST) involved in this NCMS-led project was to develop and evaluate a practical material test method suitable for dielectric permittivity of high-k polymer composite films that covers a broad frequency range including the microwave.

Currently, there are no standard test procedures for evaluating the electrical properties of high relative dielectric constant films in the frequency range of practical importance viz., at low frequencies where charge storage is important and at microwave frequencies where sub-nanosecond discharge rates are important. The current American Society for Testing and Materials (ASTM) and The Institute for Interconnecting and Packaging Electronic Circuits (IPC) test procedures are only suited for a narrow frequency range and are designed for large samples. For example, ASTM D150, D669 and D163¹ are for near 1 MHz, ASTM 3380, IPC-TM-650 No. 2.5.5.5)² are for frequencies from 8 GHz to 12 GHz. We employed several test patterns, specifically designed to evaluate the dielectric properties of high-k polymer composite films that were used by the NCMS Embedded Distributed Capacitance Project. The dielectric constant of the embedded capacitance materials was measured in the frequency range of 100 Hz to 5 GHz. The testing also included evaluation of the capacitance density, leakage current, and effect of highly accelerated temperature and humidity stress test, (HAST)³ on the capacitance. In this work, all dielectric constants are referred to the relative dielectric constant of a material.

Chapter 2

POLYMER THIN-FILM MATERIALS

Five polymer thin-film materials are discussed in this report. They were designed for the same application and prepared by microelectronic companies. All specimens are polymer composites containing high dielectric constant ferroelectric powders. Each company uses their own specific polymer and special powder to form their composite. Each company had optimized the processing conditions for the specific composite material. However, all films were processed in to the format according to our specifications for evaluation. The essential information of the polymer composite thin-films supplied by the film providers are listed in Table 1. For comparison, we have prepared a reference thin-film of FR4, Film A. FR4 is a fiberglass epoxy laminate, of which we assume the dielectric properties are quite well known. Film A has been prepared at same conditions.

Table 1
Polymer Composite Thin-Film Materials

Film	Thickness (μm)	Dielectric Constant	Material
A	100	4	FR4
B	50	3.9	Polymer composite
C	8	20	Polymer composite
D	48	11.8	Polymer composite
E	100	36	Polymer composite

Chapter 3

THE TEST VEHICLES

In this section we will describe two test vehicles in details. The term "test vehicle" is used since it is a completely processed test specimen with a test circuitry on it. Each specimen was prepared starting from a raw polymer composite material, rolled into the form of thin film, built into a multi-layer circuit board, and then patterned with a designed test circuit. A typical process will involve many detailed steps and very specific methods, which will not discuss here. Our descriptions here are limited to the design of two test vehicles. Test vehicle 0 (TV0) was used for low frequency measurement. Test vehicle 1 (TV1) was used for high frequency measurement. Two test vehicles, TV0 and TV1 were built into 3" x 5" (10 cm x 15 cm) circuit boards.

The layout of the TV0 test specimen is shown in Figure 1. It contains five patterns, *A-E*, which were designed to characterize the dielectric properties of films from DC to RF frequencies. Pattern *A* is a capacitive termination designed for the APC-7 coaxial airline at frequencies of up to 1 GHz. Patterns *B*, *C* and *D* represent a parallel plate capacitance for frequencies below 200 MHz. Pattern *E* was designed to assess electromigration resulting from the residual charge impurity levels at the surface and in the bulk of the materials. A guard surface circuitry in patterns *B*, *C* and *D* minimizes the effect of the fringing field and can be used to measure the surface leakage current. The active area, A_f of the pattern *B*, is 1.65 cm^2 (0.255 inch^2). The active area of pattern *D* is 0.79 cm^2 (0.122 inch^2), which is approximately half of the area of the *B* pattern. Pattern *C* has the back plane circuitry area and the active circuitry area perforated, $A_f = 1.58 \text{ cm}^2$, which allows for evaluating the effects of moisture absorption and/or the manufacturing process on the materials nominal capacitance density $C_d = C_m / A_f$, where C_m represents the value of the measured capacitance. The test pattern embedded inside the TV0 card can be electrically accessed via plated through holes. Characterization can also be performed on circuitized metal-clad cores to address the effects of manufacturing.

The high frequency test vehicle (TV1) shown in Figure 2 consists of microstrip lines, resonators and the corresponding co-planar terminations arranged in the three sections *A*, *B* and *C* having the line width of 400 μm , 200 μm and 125 μm respectively. The microstrip dimensions and coupling conditions were optimized to avoid overlapping of higher order modes and to achieve a low loading level at the resonance^{4, 5}. In addition to the microstrip resonators and lines for the through-reflection-line (TRL) measurements, there are five time-domain-reflectometry (TDR) sections for measuring permittivity based on a TDR response to a fast, 12.5 ps step voltage rather than on detecting the discrete, resonant frequencies. We have developed this technique specifically for thin films⁶. The TDR sections *A*, *B* and *C* contain coplanar capacitors and short terminations. The TDR section *D* is designed for the 7 mm APC standard while the dimensions of the TDR section D corresponds to the 3.5 mm APC standard, which is more suitable for films with the highest capacitance density.

Most of the difficulties in measuring the dielectric constant of high-k thin films at higher frequencies arise from the small thickness of the specimen. For such films, the desirable electrical parameters can only be realized with very small patterns of highly conducting metal traces that are difficult to fabricate and evaluate. Since the width of typical traces in plastic packages is 200 μm or larger, the characteristic impedance of such thin film devices is low. The usual calibration standards (coaxial shorts, opens and loads) become unreliable. To improve the dynamic range of the measurements, we added to the test specimen non-coaxial terminations for in-situ calibration verification.

The currently available test methods are applicable to dielectric sheets thicker than 1.2 mm with the dielectric constant lower than 12. The new microstrip test pattern is appropriate for films as thin as 40 μm having a relative dielectric constant of 50 at frequencies of up to 5 GHz. For thinner films, such as Film C, we have applied time domain reflectometry method (TDR) using corresponding test patterns to extend the frequency range of the data obtained from the microstrip resonators⁷. Unlike other techniques, TDR can provide information about the high frequency dielectric behavior from the total reflection in the presence of a large charging response associated with a large capacitance.

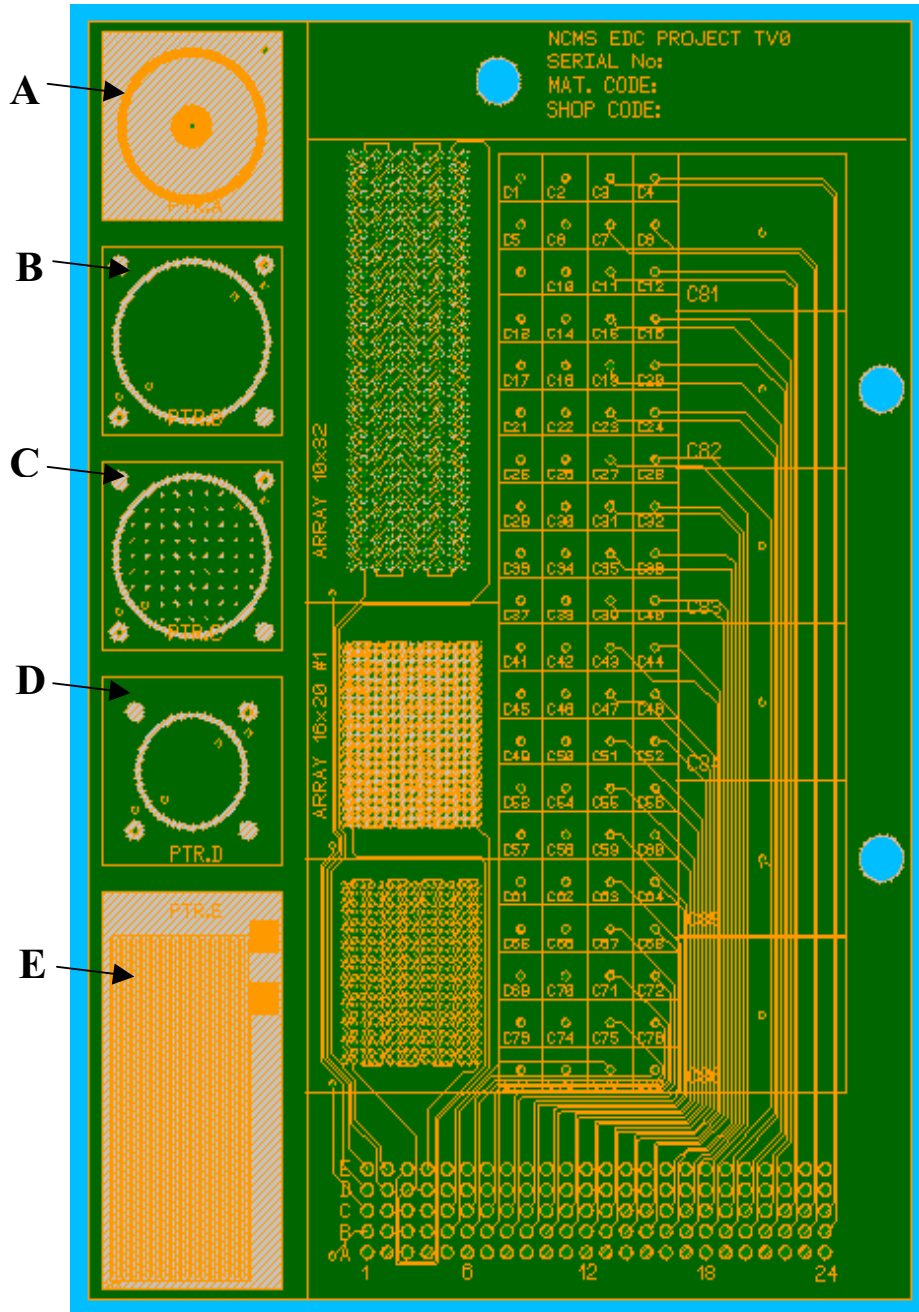


Fig. 1 Test patterns A-E for dielectric characterization of embedded capacitance films at DC and low frequencies.

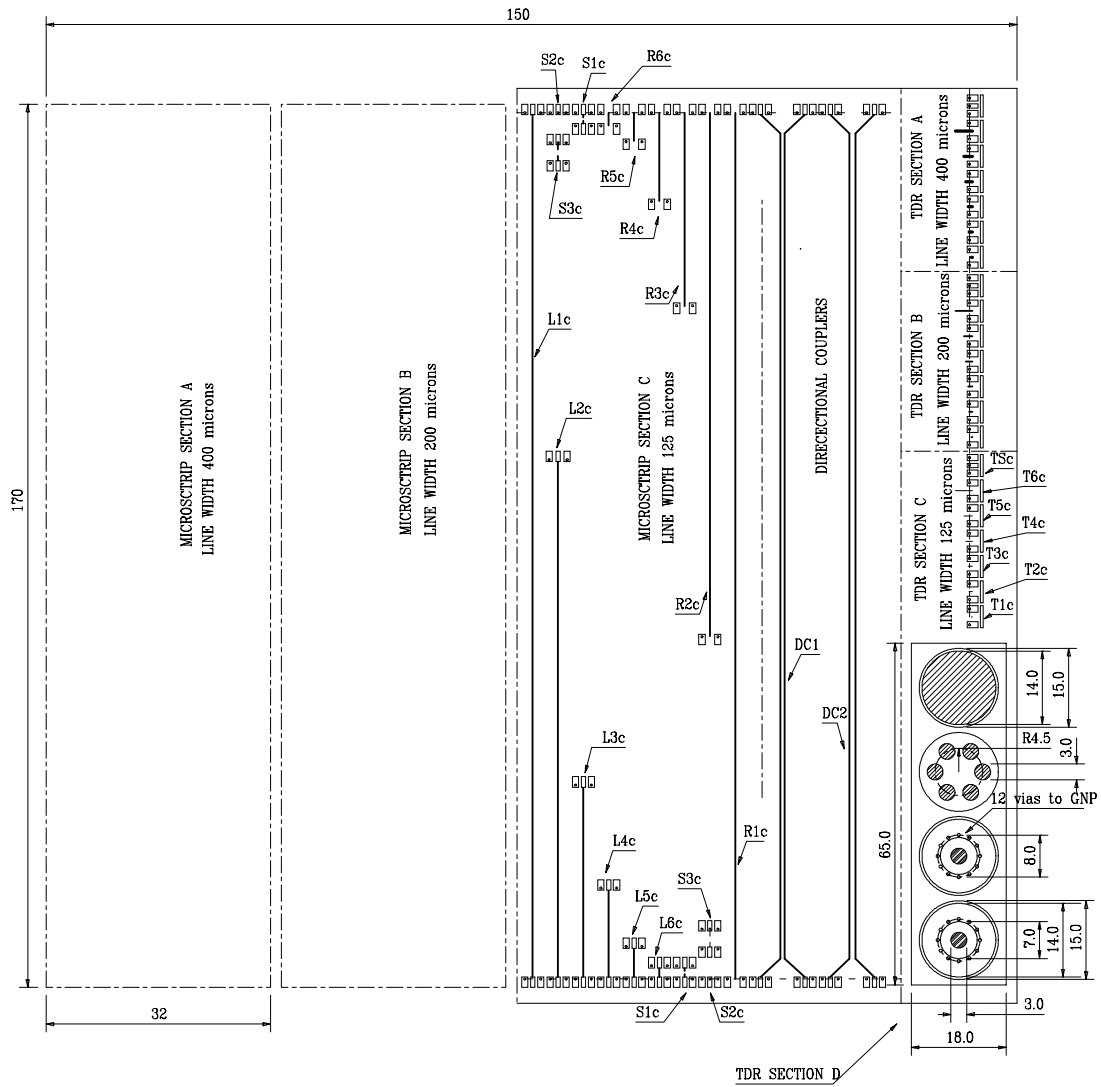


Fig. 2 Schematic of the test pattern for dielectric evaluation of high K films at microwave frequencies.

Chapter 4

EXPERIMENTAL

The low frequency measurements were carried out in accordance to the ASTM D-150 standard test method¹ using a HP4274A / HP4282A Impedance Analyzer. The measurements were performed on TV0 patterns *B* and *C* and *D*. The instrumental reproducibility of measurement of complex capacitance is less than 0.5 % of the measured value. The effect due to measurement of the fringing field was negligible since the electrical configuration included a guarding electrode. The uncertainty in the measurement of the electrode size and film thickness was estimated to be about 4 % of the data. The combined standard uncertainties of low frequency dielectric constant and $\tan \delta$ listed in Tables 1 to Table 5 are about 5 % of the reported value.

Measurements in the high frequency range were performed using the high frequency test vehicle (TV1, Figure 2) on several microstrip test specimens fabricated to fine dimensions using photolithography. The resonant frequencies were detected by measuring the $S_{2,1}$ parameters using a HP 8720D Network Analyzer and Cascade ACP-40-W GSG probes. Example results are shown in Figure 3 and Figure 4. The temperature-dependent measurements were carried out in the EC-12 Environmental Chamber from Sun Microsystems. The relative dielectric constant, ϵ_r , was calculated from the following formula⁴

$$\epsilon_r = \frac{c^2}{f_{m,n}^2} \left\{ \left(\frac{m}{2l} \right)^2 + \left(\frac{n}{2w} \right)^2 \right\}$$

where l and w are the electrical length and width of the resonator, m and n are the mode integers. The dielectric loss was estimated from the width of the resonance peak, measured 3 dB below its maximum value. The largest contribution to the uncertainty in this method comes from the coupling length, which affects the total length of the resonator. The

combined standard uncertainty of high frequency dielectric constant and $\tan \delta$ listed In Table 1 to Table 5 increases from about 2 % at 1 GHz, to 8 % of the value at 5 GHz. The capacitance and $\tan \delta$ data shown in the form of figure in the result section has the same uncertainty as discussed here.

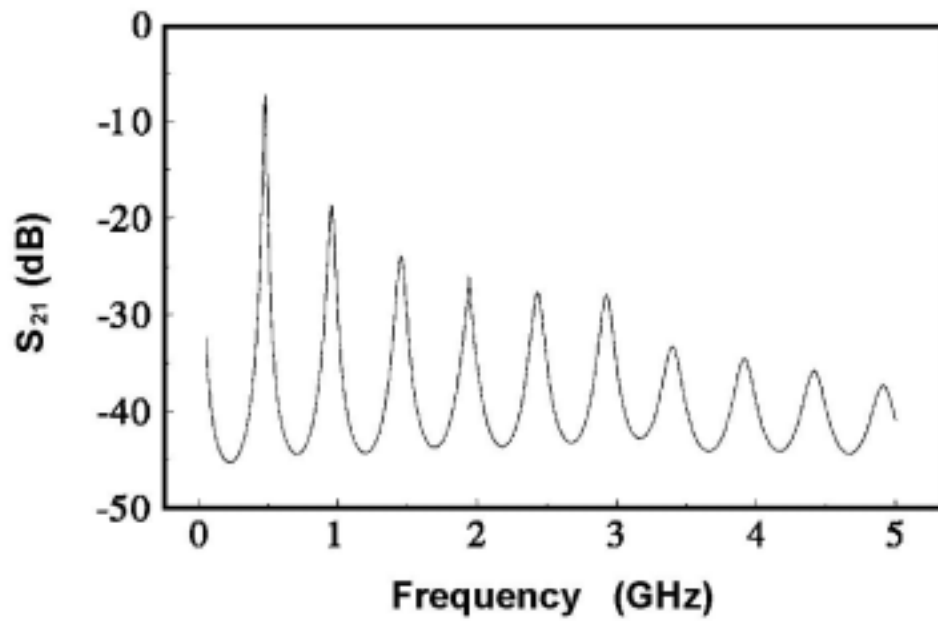


Figure 3 Measurement example using the microstrip test of Film B.
(the resonator length $l=164$ mm, $\epsilon^1 = 3.93$ at 461 MHz).

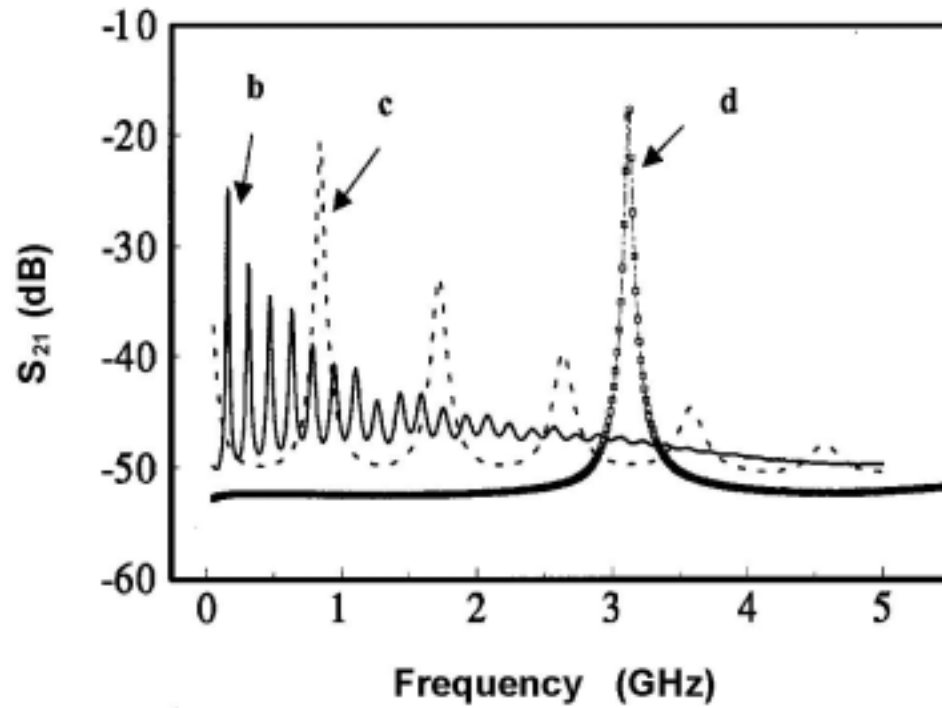


Figure 4 Measurement example using the microstrip test of Film E. b) $l=164$ mm, $\epsilon^1 = 36.3$ at 151 MHz; c) $l=29.5$ mm, $\epsilon^1 = 36.4$ at 842 MHz; d) $l=8$ mm, $\epsilon^1 = 35.9$ at 3.122 GHz.

Chapter 5

RESULTS AND DISCUSSION

The five film capacitance materials listed in Chapter 2 were evaluated. For comparison, data was interpreted in terms of that obtained for the FR4 fiberglass epoxy laminate in analogous conditions. The low frequency measurements were performed on TV0 pattern *B* and *C*.

Polymer Film A

Figure 5 shows the DC insulation resistance of FR4 material under 100 V bias as a function of temperature. The insulation resistance decreases from about $1 \times 10^{11} \Omega$ at 85 °C to $5 \times 10^7 \Omega$ at 180 °C, indicating an excellent electrical insulation characteristic.

The capacitance density and the dielectric loss tangent of the FR4 material as a function of frequency and temperature are shown in Figure 6 and Figure 7, respectively. In the glassy state, below 125 °C, FR4 shows a relatively insignificant dependence of the capacitance on frequency, which reflects the behavior of the dielectric constant of the material (referred as dielectric dispersion). Following the dielectric dispersion, the capacitance density decreases slightly with increasing frequency. In the temperature range of 20 °C to about 130 °C, the capacitance density is approximately 0.04 nF/cm^2 . Above 130 °C, the FR4 resin enters the rubbery state and the capacitance increases considerably due to the liberation of large-scale motions of the resin backbone. This molecular behavior is also responsible for an increase in the dielectric loss tangent at temperatures approaching the glass-rubber transition. It is noteworthy that the low frequency loss tangent measurements detect the beginning of the glass-rubber relaxation process at temperature typically (10 to 15) °C below the T_g determined by the DSC or TMA techniques. The 200 Hz curve in Figure 7 exhibits a low temperature shoulder of the α -relaxation peak, which shifts to higher temperatures as the

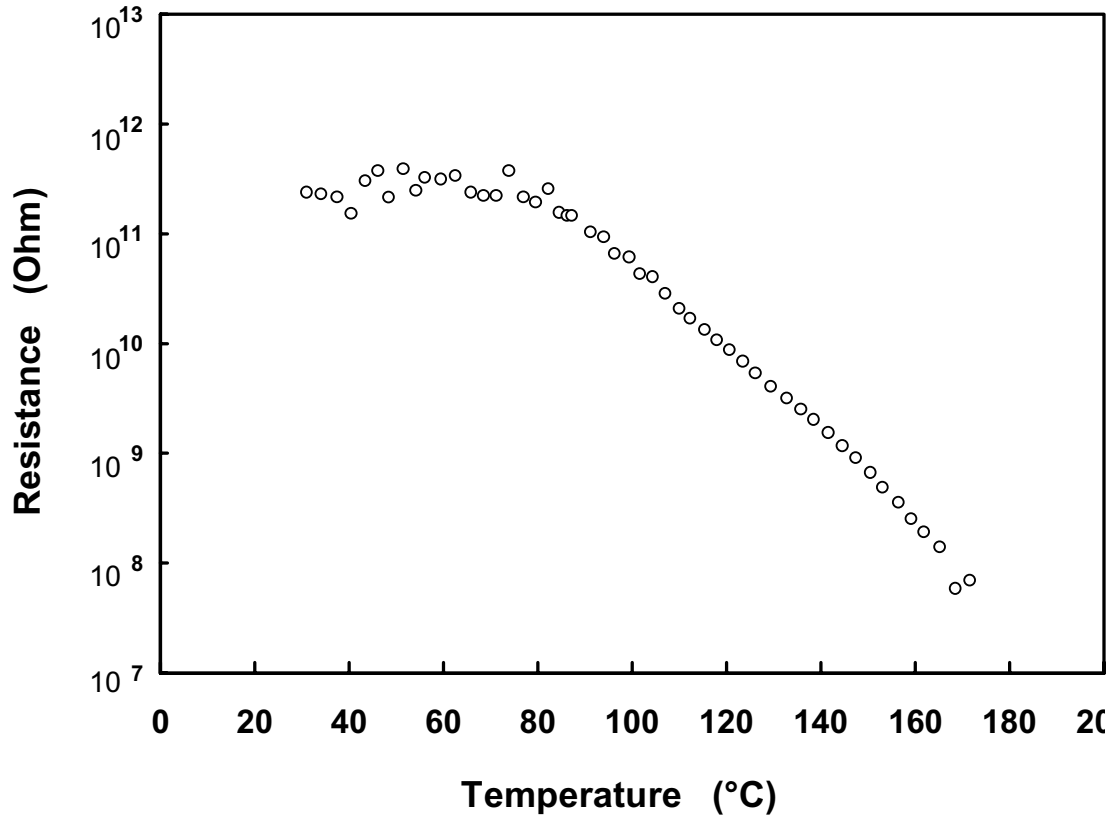


Figure 5 Insulation resistance of Film A under bias of 100V as a function of temperature.

frequency increases. In other words, the dielectric loss at elevated temperatures appears to decrease as the frequency increases.

In addition to the α relaxation, FR4 also exhibits a local β -relaxation, which strongly affects the dielectric performance of the material in the RF and microwave frequency range. A high temperature shoulder of the β -relaxation peak is seen on the 100 kHz curve in Figure 7 as a decrease in the dielectric loss in the temperature range of 20 °C to about 140 °C. This peak shifts to lower temperatures with decreasing frequency. Therefore, the room temperature dielectric loss in Figure 7 appears to increase with increasing frequency. For a typical FR4 resin, the room temperature dielectric loss tangent reaches a maximum of about 0.025 at 10 MHz. The loss due to β relaxation should decrease at higher frequencies (room temperature) though molecular defects, free radicals and other paramagnetic impurities may contribute to the dielectric loss mechanism, extending it into the microwave. The dielectric constant data in the high frequency range obtained for FR4 using the microstrip test pattern is given in Table 2.

Moisture absorbed during a HAST test increases the dielectric loss and the dielectric constant (capacitance density). The largest effect is seen at the lowest frequencies (Fig 8) due to interfacial polarization. At 200 Hz, the change in capacitance after HAST is about 6 pF/cm². In comparison, the dielectric loss tangent increases considerably higher: from 7×10^{-3} to about 2×10^{-2} . This indicates that the absorbed water may facilitate electromigration and compromise the reliability of the dielectric under DC bias or when carrying a digital ON/OFF signal.

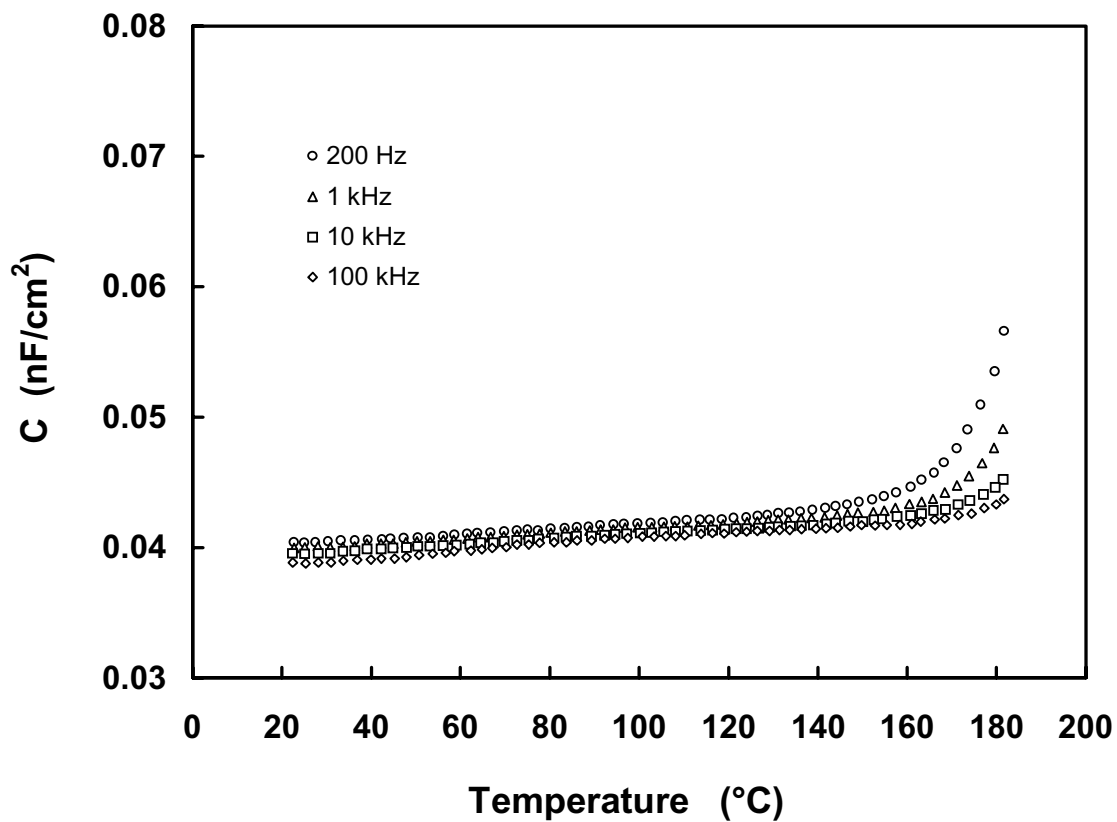


Figure 6 Capacitance density of Film A as a function of temperature and frequency. (TV0 /Metrics, pattern B)

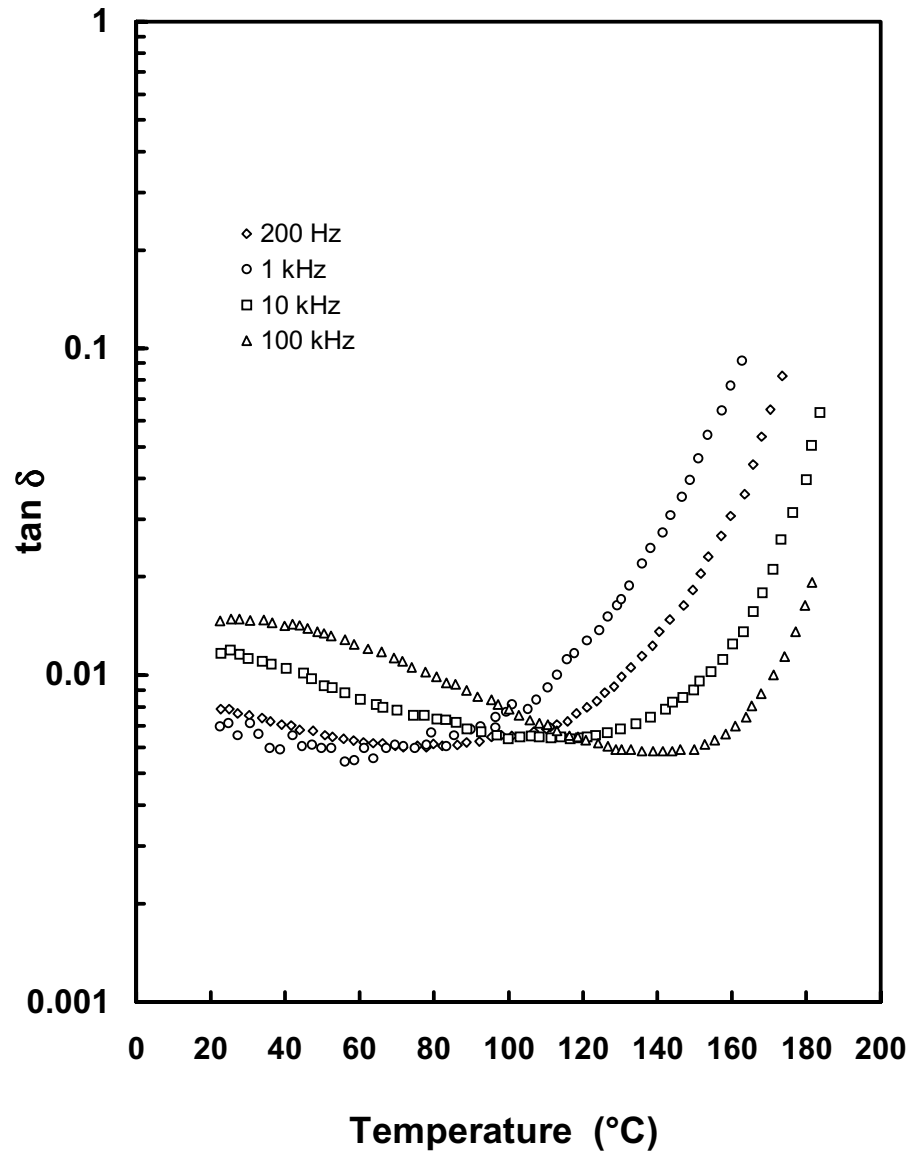


Figure 7 Dielectric loss tangent of Film A as a function of temperature and frequency.(TV0 / Metrics, pattern B)

Table 2

High Frequency Dielectric Constant of Film A

Frequency (GHz)	Dielectric Constant	Loss Tangent
0.450	4.13	0.023
0.924	3.92	0.021
1.398	3.85	0.02
1.874	3.81	0.018
2.348	3.79	0.017
2.822	3.78	0.017
3.290	3.78	0.015
3.771	3.78	0.014
4.226	3.79	0.012

(resonator length $l=164$ mm)

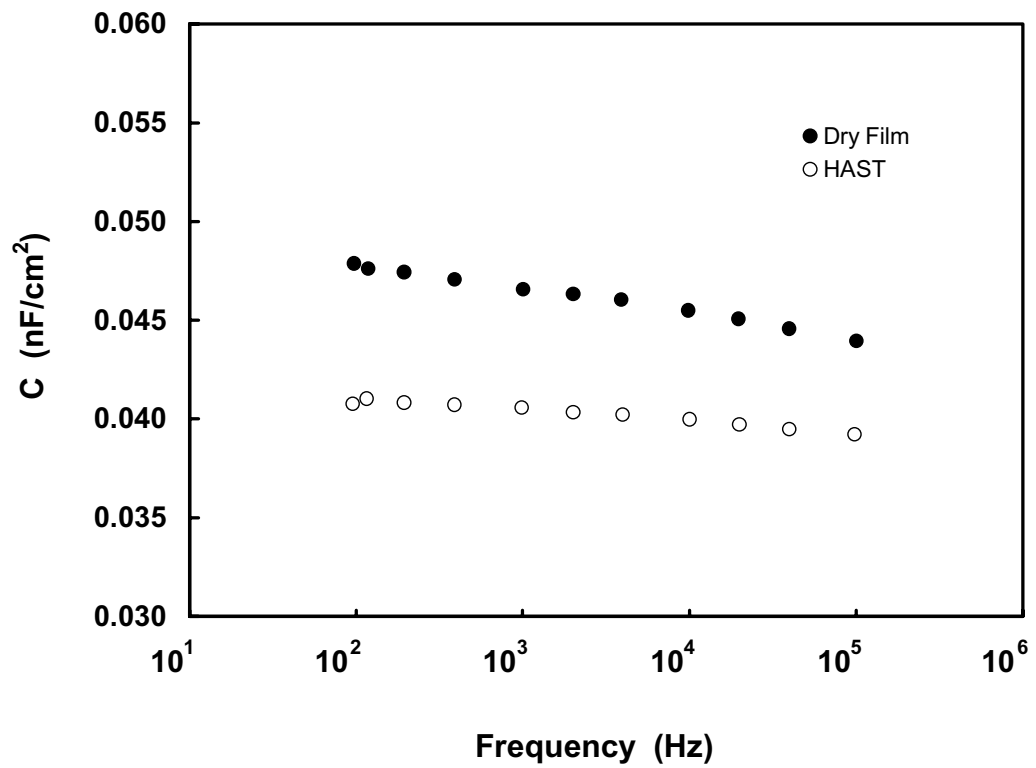


Figure 8 Capacitance density of Film A saturated with moisture during HAST. (TV0 / Metrics, pattern C)

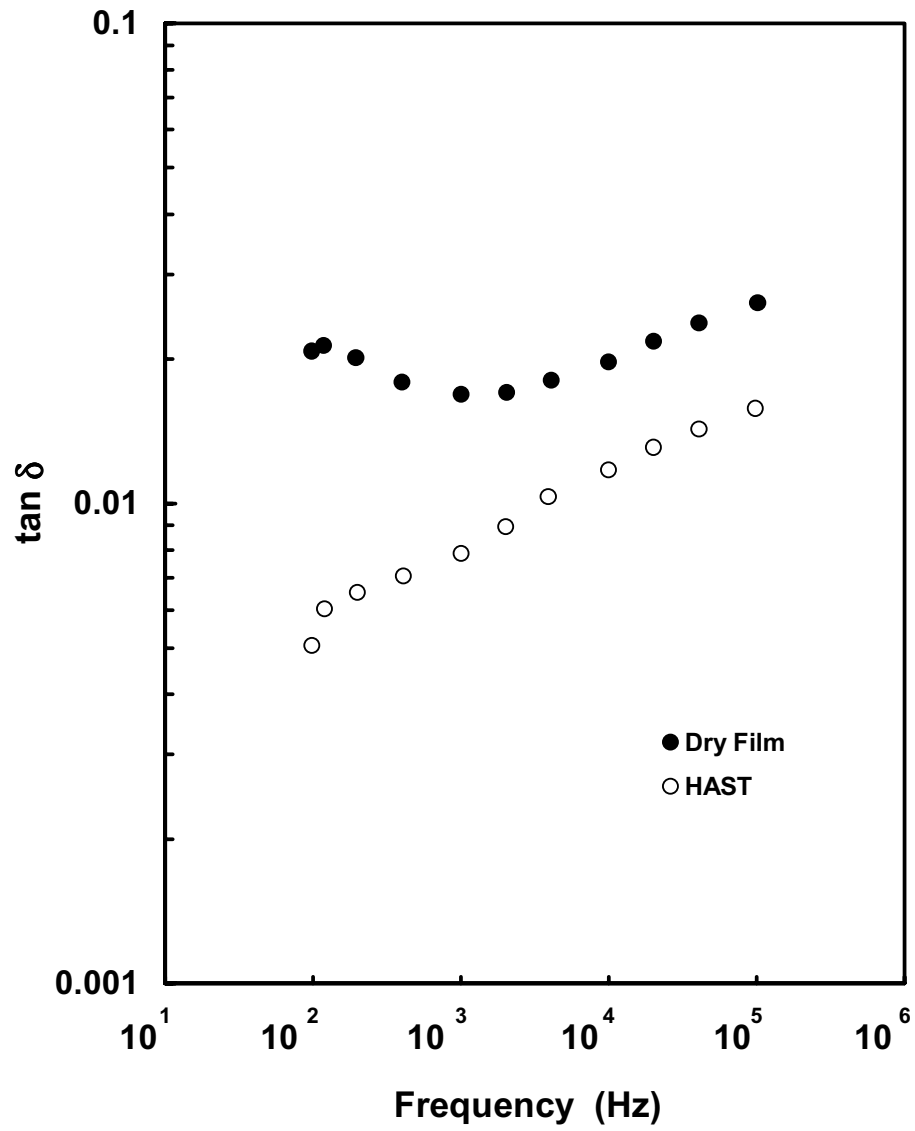


Figure 9 Dielectric loss tangent of FR4 saturated with moisture d during HAST. (TV0 / Metrics, pattern C).

POLYMER FILM B

The DC insulation resistance of the Film B shows a thermally activated behavior similar to that of FR4 (Fig 5). The insulation resistance decreases from about $1 \times 10^{11} \Omega$ at $85 \text{ }^\circ\text{C}$ to $1 \times 10^7 \Omega$ at $180 \text{ }^\circ\text{C}$. A change in the activation energy of the conducting process takes place at about $125 \text{ }^\circ\text{C}$ due to glass-rubber transition of the resin. The capacitance density is about 0.08 nF/cm^2 , (0.5 nF/inch^2) approximately twice that observed in the FR4. This is due to the decrease in thickness of the dielectric. The capacitance density and the dielectric loss tangent plots shown in Figures 10 and 11 for Film B as a function of frequency and temperature are analogous to that of FR4 (Figures 6 and 7).

The dependence of capacitance density (the dielectric constant) on temperature is somewhat stronger than that of FR4, probably due to lower glass transition temperature. Consequently, both the large-sack and local relaxation are more temperature dependent, which contributes to an increase in loss at the RF and microwave range. The dielectric constant data in the high frequency range obtained for Film B using the microstrip test pattern is given in Table 3.

Saturation with moisture during HAST increases capacitance density by about 3 pF/cm^2 from 0.081 nF/cm^2 to 0.084 nF/cm^2 . Moisture also contributes to an increase in the dielectric loss. The large effects were similar to those shown in Figs. 8 and 9 for FR4 at the lowest frequencies.

Table 3

High frequency dielectric constant and
loss tangent of Film B

Frequency (Ghz)	Dielectric Constant	Loss Tangent
0.461	3.93	0.024
0.928	3.88	0.021
1.399	3.84	0.021
1.872	3.82	0.020
2.347	3.8	0.019
2.820	3.78	0.020
3.291	3.78	0.016
3.774	3.75	0.016
4.225	3.79	0.015

(Resonator length $l=164$ mm)

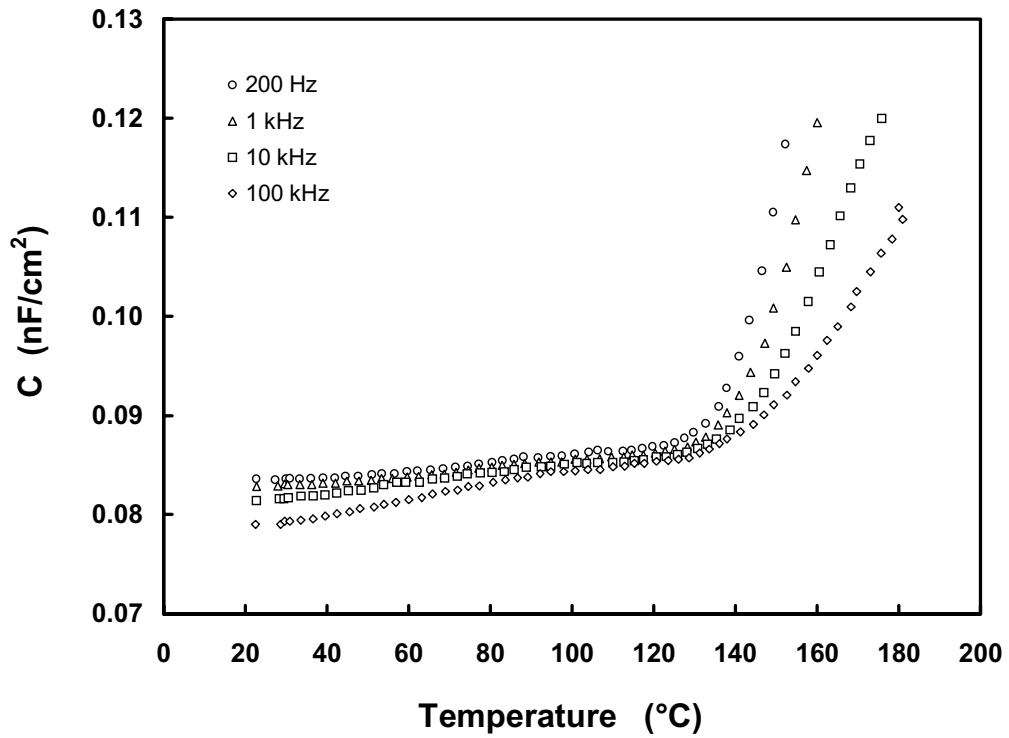


Fig.10 Capacitance density of Film B as a function of temperature and frequency (TV0, pattern B)

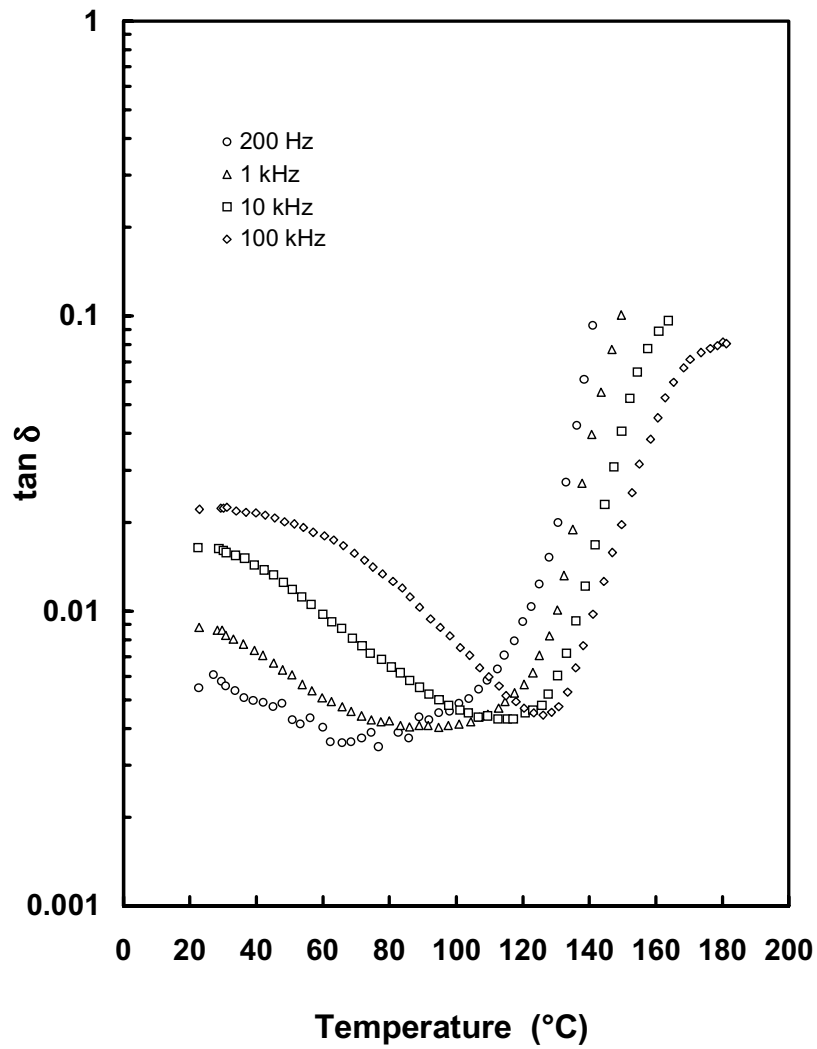


Fig.11. Dielectric loss tangent of Film B as a function of temperature and frequency.(TV0, pattern B)

Polymer Film C

The capacitance density of the Film C, 8 μm thick films were determined to be about 2.5 nF/cm² (16 nF/inch²) at ambient conditions. The material shows a depressive behavior, similar to that of epoxy-based composites. The capacitance density and the dielectric loss tangent plots are shown in Figures 12 and 13 as a function of temperature at several frequencies. A thermally activated increase in capacitance is seen above 85 °C (Figure 12).

The data in Figure 12 and 13 is reminiscent of a resin blend that has a glass transition temperature below 130 °C. Therefore, the temperature dependence of capacitance density (dielectric constant) is stronger than of FR4 and both the large sack and local relaxation are more temperature dependent. This is evidenced as an increase in the dielectric loss. The dielectric constant at high frequencies, above 1 MHz, was measured using the TDR technique (Table 4). Due to the small thickness and large capacitance, the 400 μm wide microstrip lines exhibited very low impedance, below 2 Ω , causing a large return loss (S_{11}) at the input reference plane. Therefore, the resonance peaks were diminished and analysis difficult, especially at frequencies above 1 GHz. In contrast, the TDR method was more suitable for determination of the dielectric constant in the presence of large charging currents. The results obtained at the ambient temperature are listed in Table 4. It is seen that Film C exhibits relatively low dispersion. The dielectric constant decreases slightly from 23 at 200 Hz to 20.5 at 2 GHz. The dielectric loss tangent increases from 0.005 to about 0.14 in the same frequency range. At 1 GHz, Film C exhibited the largest loss among the investigated materials.

Table 4

High frequency dielectric constant and
loss tangent of Film C

Frequency	Dielectric Constant	tan(δ)
200 Hz [*]	23.2	0.00572
1 kHz [*]	23.1	0.008510
100 kHz [*]	22.5	0.0135
1 MHz ^{**}	22.4	0.0250
10 MHz ^{**}	22.0	0.0387
100 MHz ^{**}	21.5	0.0436
1.0 GHz ^{**}	21.1	0.0810
1.5 GHz ^{**}	20.8	0.133
2.0 GHz ^{**}	20.5	0.141
2.5 GHz ^{**}	20.5	0.145

* LCR Impedance Analyzer, TV0 pattern B.

** HF TV, TDR pattern D

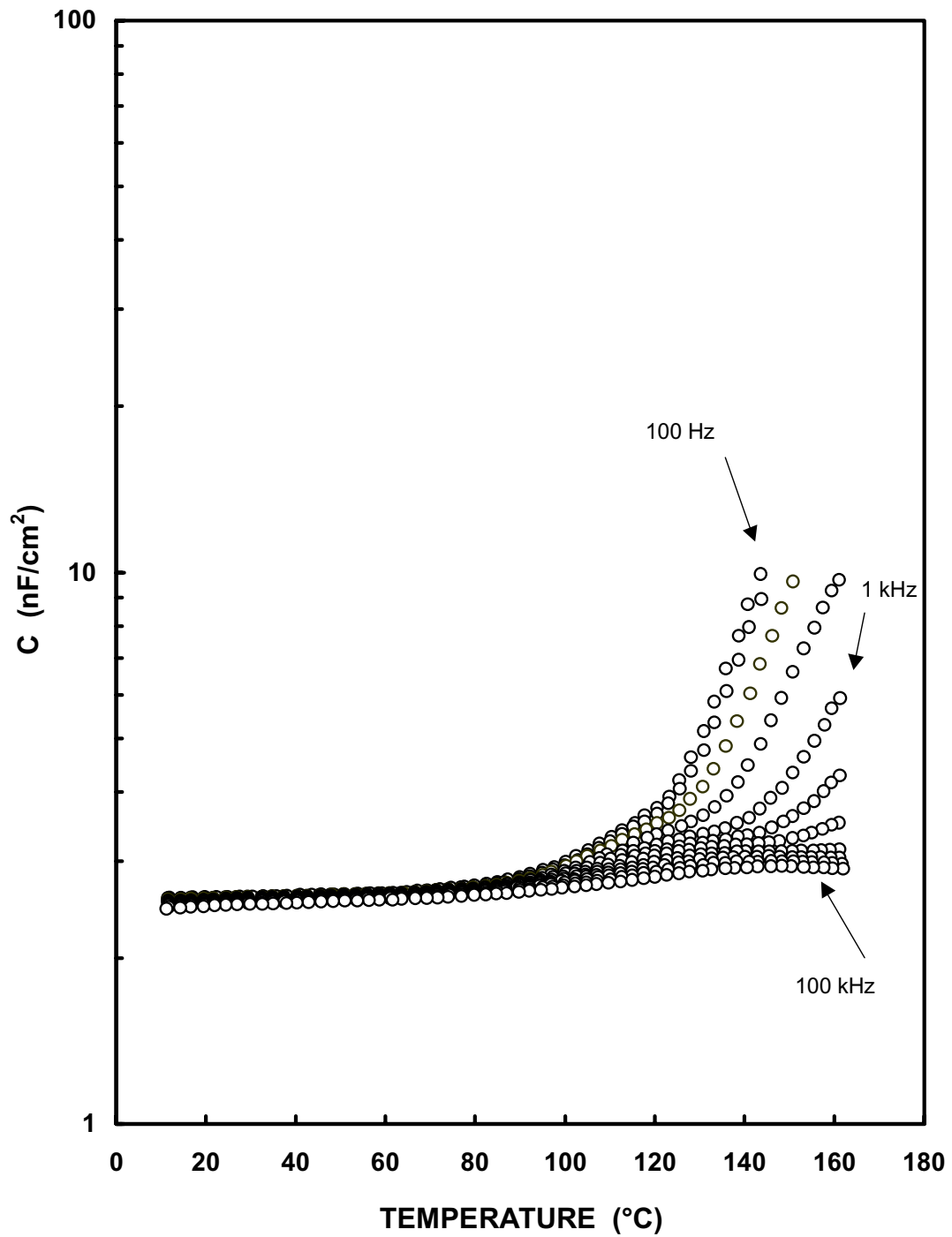


Fig. 12 Capacitance density of Film C as a function of temperature and frequency. (TV0, pattern B)

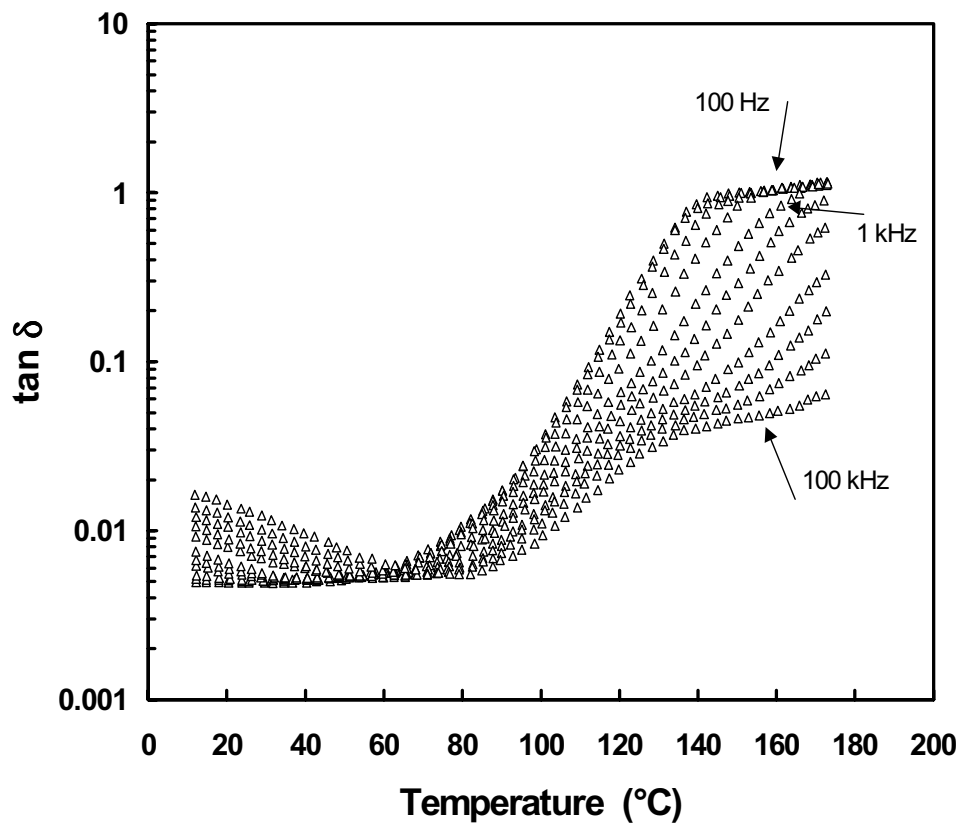


Fig.13 Dielectric loss tangent of Film C as a function of temperature and frequency. (TV0, pattern B)

Polymer Film D

The insulation resistance of Film D decreases with increasing temperature exponentially from about $3 \times 10^9 \Omega$ at 85°C to $5 \times 10^6 \Omega$ at 180°C . In contrast to the epoxy-based composites, which undergo a glass-rubber transition at about 130°C , the thermally activated DC conductivity of Film D in the glassy state has single activation energy. The capacitance density is about 0.25 nF/cm^2 , which corresponds to the dielectric constant 12 for films of $41 \mu\text{m}$ in thickness at ambient temperatures.

Figures 14 and 15 shows the effect of temperature on the capacitance density and the dielectric loss tangent at several frequencies from 200 Hz to 100 kHz. At ambient conditions the capacitance density is about 0.25 nF/cm^2 (1.5 nF/inch^2) without noticeable dispersion, i.e. no frequency dependence. The capacitance density increases with temperature only at frequencies below 10 kHz. This behavior probably originates from the interfacial arrangement and polarization between separate phases in the composite rather than from polarization of the polyimide backbone itself. It is seen that at higher frequencies, above 10 kHz, the capacitance density appears to be independent of temperature. This is in contrast to the epoxy-based composites, which typically show thermally activated increase in capacitance when the rubbery transition of the resin is approached. It is worthy to note that the dielectric loss tangent decreases with increasing frequency.

At room temperature, Figure 15, the dielectric loss tangent decreases from approximately 0.01 at 200 Hz to 0.002 at 1 MHz (Table 5). In comparison, the epoxy-based composite films show an increasing loss at similar frequencies and temperatures. The dielectric data of high-k polyimide at higher frequencies is listed in Table 5 where the dielectric constant decreases from 12.5 at 200 Hz to about 11.6 at 6 GHz. An apparent increase in the dielectric loss tangent at frequencies of 265 MHz and above may be the result of electrical limitations of the microstrip test specimen (limited Q factor) at higher frequencies.

Saturation with moisture during HAST increases capacitance density and loss of Film D considerably. The effect is especially large at the lowest frequencies and might be due to interfacial polarization and accumulation of charge carriers between the polyimide resin and the ceramic phases. At 200 Hz, 25 °C, the difference in capacitance density between the dry and moisture-saturated samples was about 120 pF/cm².

Table 5

High frequency dielectric constant and
loss tangent of Film D

Frequency	Dielectric Constant	Loss tangent
200 Hz	12.5	0.01
1 kHz	12.3	0.0048
10 kHz	12.2	0.0032
100 kHz	12.1	0.0031
1 MHz	11.9	0.0029
265 MHz	11.9	0.0082
757 MHz	12.1	0.011
1.0931 GHz	11.8	0.0098
3.0278 GHz	11.7	0.0122
3.6511GHz	11.7	0.011
5.9642GHz	11.6	0.011

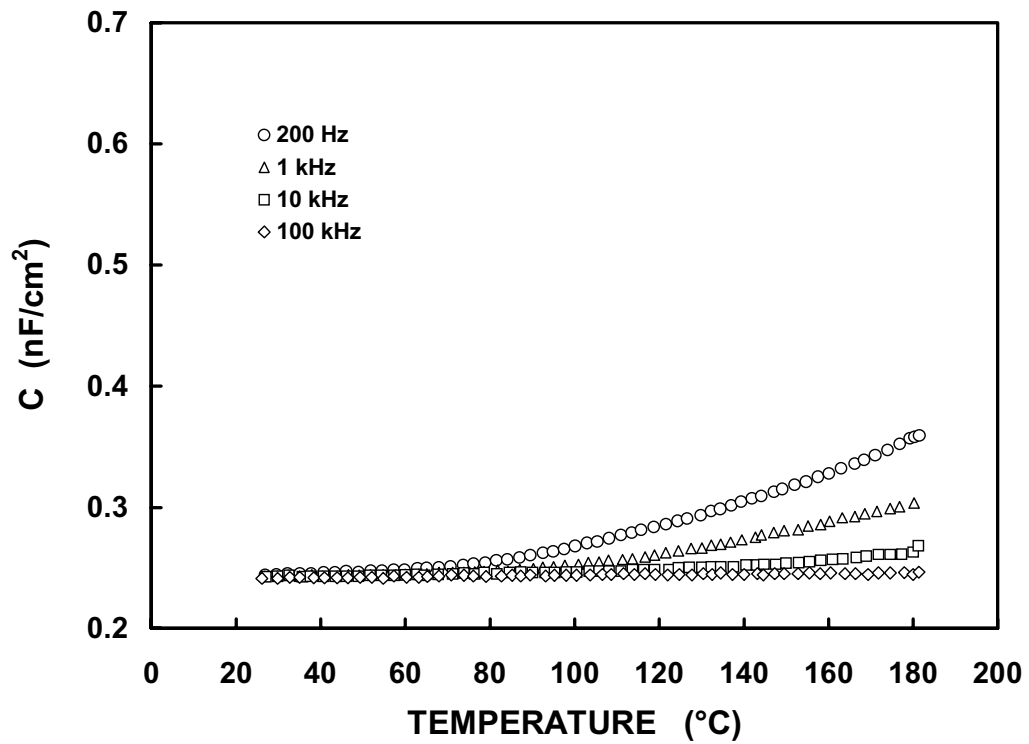


Fig.14 Capacitance density of Film D as a function of temperature and frequency.(TV0, pattern B)

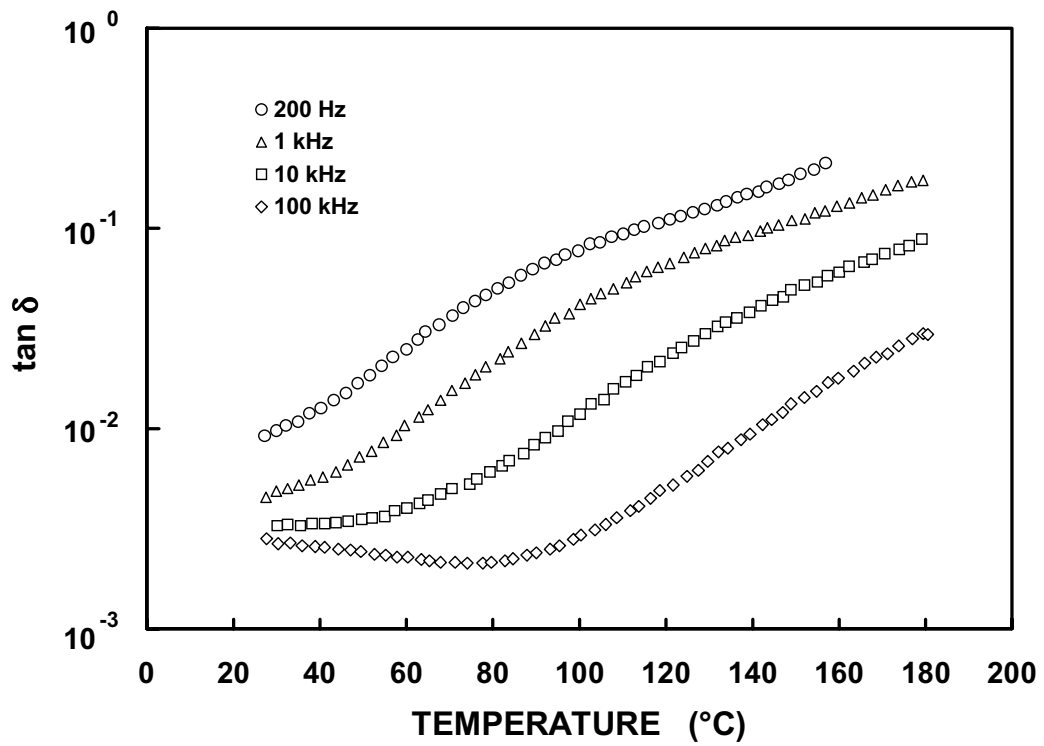


Fig.15 Dielectric loss tangent of Film D as a function of temperature and frequency. (TV0, pattern B)

Polymer Film E

The DC insulation resistance of Film E shows a thermally activated behavior similar to that seen in Figure 5 for FR4. The insulation resistance decreases from about $10^{11} \Omega$ at 85 °C to $10^7 \Omega$ at 180 °C. A change in the activation energy of the conducting process was detected at about 125 °C, which originated from the glass-rubber transition of the resin matrix.

The capacitance density measured at ambient temperature was about 0.33 nF/cm^2 (2.1 nF/inch^2), which for films 100 μm thick corresponds to a dielectric constant of about 37. This is one of the highest dielectric constant values measured for an epoxy composite loaded with ferroelectric powder. Since the practically achievable loading level is a volume fraction of about 50 % of the ceramic powder phase, the BaTiO_3 powder-epoxy composites may exhibit a dielectric constant about 50. The capacitance density and the dielectric loss tangent plots as a function of temperature at several frequencies are shown in Figure 16 and 17. A thermally activated increase in capacitance is seen above 130 °C (Figure 16). The character of the dielectric loss data is reminiscent of the FR4 epoxy resin laminates. This indicates that the organic matrix rather than the high dielectric constant ceramic filler dictate the dielectric loss. Thus, an increase in the dielectric loss with increasing frequency seen at room temperature originates from the relaxation processes in the polymer.

Table 6 lists the room temperature dielectric constant and loss obtained for Film E using the low frequency pattern B (200 Hz to 1 MHz) and the microstrip test pattern (100 MHz to 8 GHz). Saturation with moisture during the 200 h HAST increases the capacitance density. At 200 Hz and 25 °C, the difference in capacitance density between the dry and moisture-saturated samples was measured to be about 70 pF/cm^2 . Moisture absorption also contributed to an increase in the dielectric loss. The effect was somewhat larger than that shown in Figs. 8 and 9 for FR4.

Table 6

Dielectric Constant and Loss Tangent of Film E

Frequency	Dielectric Constant	Loss tangent
200 Hz	39.2	0.00774
1 kHz	38.8	0.0101
10 kHz	38.2	0.0151
100 kHz	37.9	0.0198
1 MHz	37.2	0.0212
99.98 MHz	36.6	0.0182
151.87 MHz	36.3	0.0168
414.32 MHz	36.3	0.0151
842.00 MHz	36.4	0.0132
1.729 GHz	36.0	0.0123
3.122 GHz	35.9	0.0114
7.921 GHz	36.0	0.01029

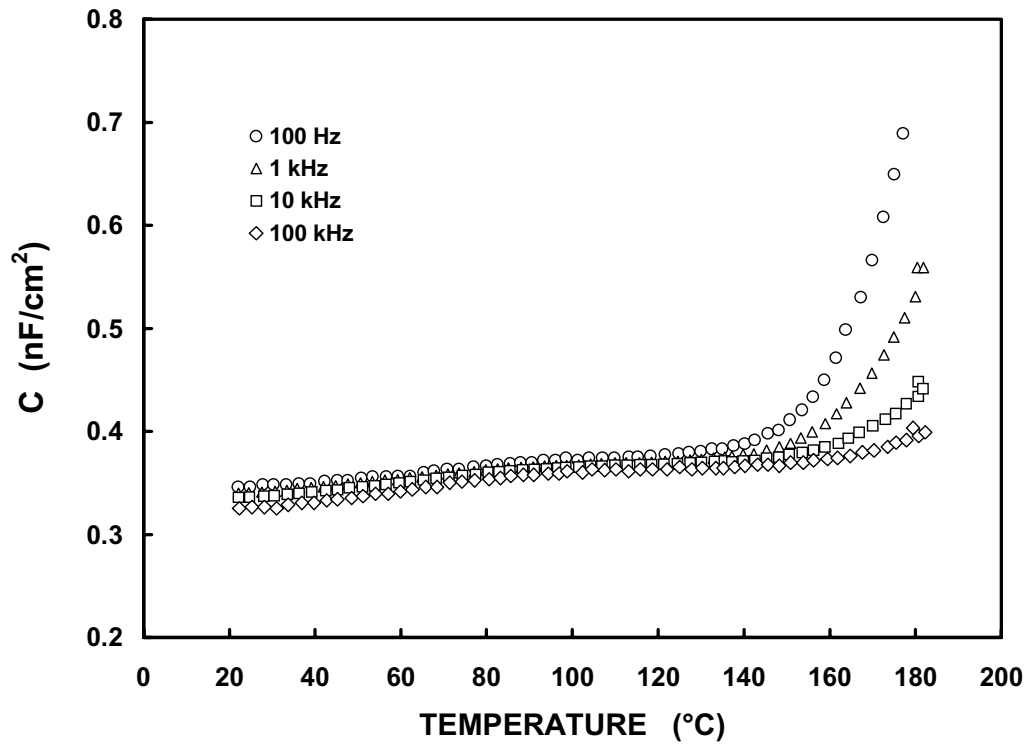


Fig.16 Capacitance density of Film E as a function of temperature and frequency. (TV0, pattern B)

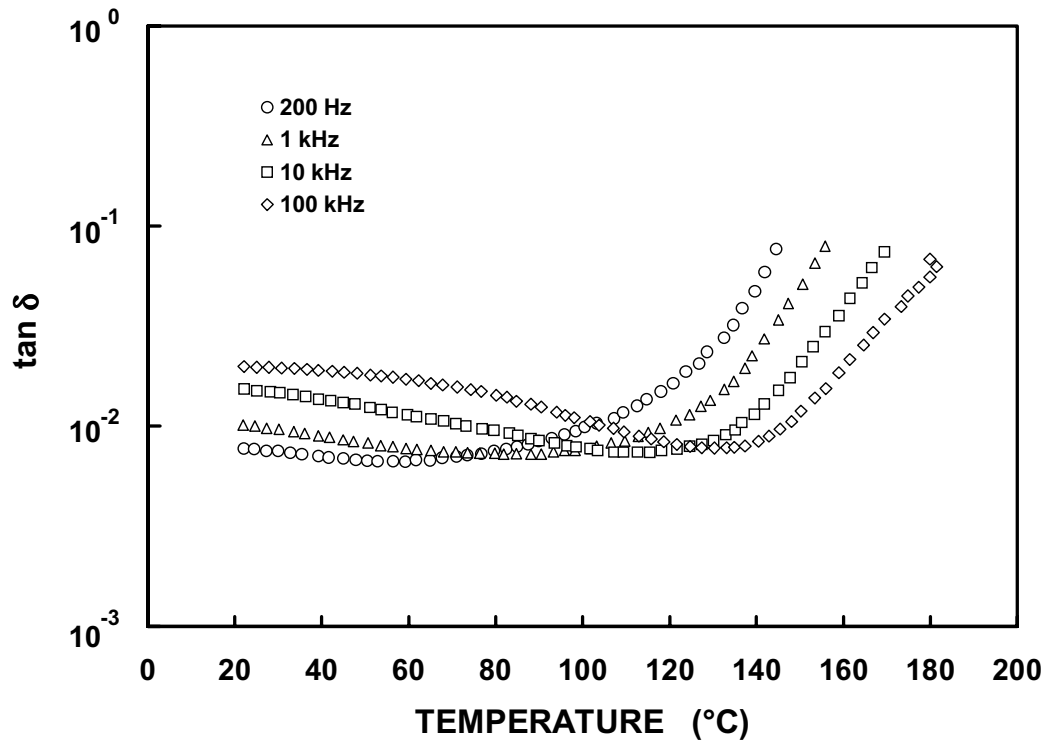


Fig.17 Dielectric loss tangent of Film E as a function of temperature and frequency. (TV0, pattern B)

Chapter 6

SUMMARY

Dielectric characterization of embedded capacitance materials has been done as a function of temperature at frequencies ranging from 200 Hz to 5 GHz. The measurements of the dielectric constant were performed in a practical, functional, thin film configuration using test specimens specifically designed for the NCMS-led Embedded Capacitance Project. The test specimens were fabricated using standard copper circuitize and card lamination processes by participant companies.

The largest dielectric constant was measured for Film E. At 1 GHz this material showed a dielectric constant of 36. The capacitance density of a 100 μm thick Film E was about 0.33 nF/cm^2 (2.1 nF/inch^2). Film C showed at 1 GHz a dielectric constant of 20. In comparison, the capacitance density of the 8 μm thick Film C was about 2.5 nF/cm^2 (16 nF/inch^2) nearly 8 times larger than that of Film E due to much smaller thickness of the dielectric film. At frequency of 1 GHz, the dielectric constant of the Film D was about 11.8, while the capacitance density of the 48 μm thick film was comparable to that of Film E, about 0.25 nF/cm^2 (1.5 nF/inch^2). Film B exhibited the lowest dielectric constant, 3.9 and the lowest capacitance density 0.08 nF/cm^2 (0.5 nF/inch^2) among the evaluated materials for 50 μm -thick films. Using the capacitance density of FR4 laminate as a reference, we arranged the tested materials in the order of increasing the capacitance density as shown in Figure 18. Our conclusion is that the capacitance density of the Film C was 63 times larger than that of FR4. This is the state of art high dielectric constant film can be made by industries if a material required today.

During heating, Films E, C and B showed a thermally activated increase in capacitance and dielectric loss, similar to that of FR4 epoxy resin when the glass transition temperature is approached. In contrast, permittivity of the Film D was much less temperature dependent, especially at higher frequencies.

The dielectric properties of all the tested thin film capacitance composites are sensitive to moisture. Water accumulated during HAST typically increase the capacitance density and the dielectric loss tangent. The largest effect was observed at lowest frequencies due to moisture-activated conductivity and interfacial polarization.

The high frequency microstrip test pattern with 400 μm wide traces was appropriate for testing high-k films thicker than 40 μm at frequencies of up to 5 GHz. We are investigating the applicability of higher impedance, narrower traces (250 μm and 125 μm wide) for films thinner than 40 μm .

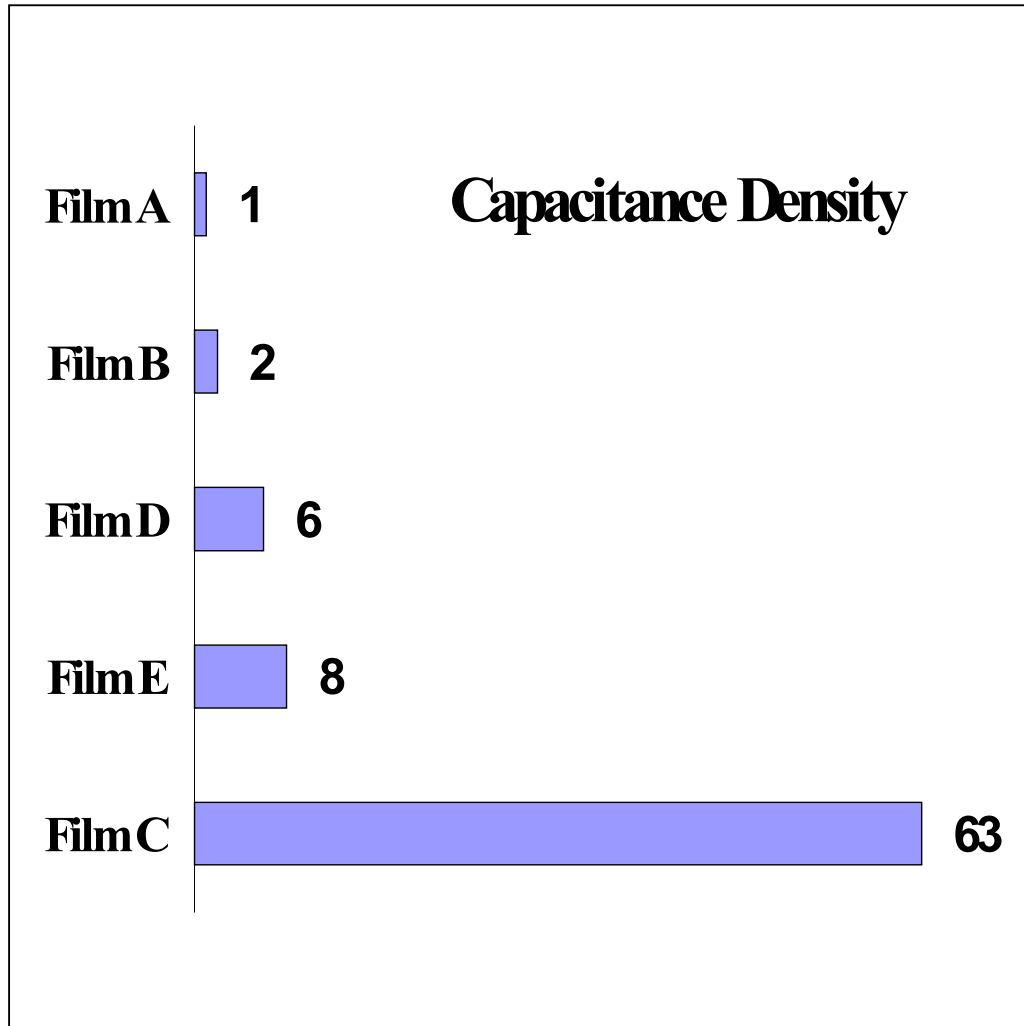


Fig.18 Comparison of microwave dielectric thin films. Capacitance density of Film A is assigned as one and used as the reference.

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DISCLAIMER

Certain materials and equipment identified in this manuscript are solely for specifying the experimental procedures and do not imply endorsement by NIST or that they are necessarily the best for these purposes.

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