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AN NMOS TEST CHIP FOR A COURSE IN SEMICONDUCTOR PARAMETER MEASUREMENTS NATIONAL EUREAU OF STANDARDS LIBRARY

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K. P. Roenker and L. W. Linholm

U.S. DEPARTMENT OF COMMERCE National Bureau of Standards National Engineering Laboratory Center for Electronics and Electrical Engineering Semiconductor Devices and Circuits Division Washington, DC 20234

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An NMOS Test Chip for a Course in Semiconductor Parameter Measurements

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ABSTRACT

This report describes an NMOS test chip, NBS-40, which was developed to be used in graduate level electronics engineering courses involving semiconductor parameter measurements associated with the fabrication of integrated circuits. The 35 test structures included in the test chip and their use in materials, device, and process parameter measurements are described. Details of the silicon gate NMOS process used in the chip fabrication are also provided.

Key words: integrated circuit; microelectronics; parameter measurements; solid-state electronics; test chip; test pattern; test structures.

1. INTRODUCTION

Semiconductor parameter measurements are used for the characterization of materials and devices, extraction of parameters for device and circuit simulation, and characterization and monitoring of integrated circuit processes associated with the fabrication of integrated circuits (ICs). The objective of this report is to describe an NMOS test chip which was developed to be used in electronics engineering courses involving semiconductor parameter measurements [1]. The test chip layout, the NMOS fabrication process used in its preparation, and the test structures and associated parameter measurements are described in this report.

2. TEST CHIP LAYOUT AND NMOS PROCESS

NBS-40 is a $3450-\mu m$ by $2760-\mu m$ test chip for a polysilicon gate NMOS fabrication process. The test chip contains 35 test structures which are configured in a 2 by 16 probe pad arrangement [2] with $80-\mu m$ probe pads located on 160- μm centers. Individual test structures are placed between the pads in the 2 by 16 column. The only exceptions to this placement are the MOS capacitor, ring oscillator, and inverter. The layout scheme permits convenient probing with one fixed probe card on a wafer prober, which may be computer controlled. The MOS capacitor is designed with larger probe pads for manual probing at a separate probe station. A complete list of test structures is given in the table; a layout of the chip is seen in figure 1. The test chip was designed to be fabricated with a nine-photomask, polysilicon gate, self-aligned NMOS fabrication process. Three ion-implantation steps are used to adjust the threshold voltage of the depletion and enhancement mode Metal Oxide Semiconductor Field Effect Transistors (MOSFETs) and to form the sources and drains. In general, a minimum device design feature size of 5 µm was used. The NMOS process described in this report is similar to that employed elsewhere [3].

A cross section and top view of a MOSFET is seen in figure 2. To produce self-aligned MOSFETs and minimize overlap capacitance, a phosphorus sourcedrain diffusion is performed using an oversize polysilicon gate. A subsequent photomask defines the final gate size, and a phosphorus implant performed resulting in self-alignment and the desired shallow junction depth $(0.3 \ \mu m)$. Details of the fabrication process are provided in the appendix.

3. TEST STRUCTURE DESCRIPTIONS

A. Cross-Bridge Sheet Resistors

Two phosphorus-diffused, one metal, and two polysilicon cross-bridge sheet resistor test structures are included in NBS-40 for the measurement of the sheet resistance and the determination of electrical linewidth for each of the conducting layers. The cross-bridge test structure, seen in figures 3a-3c, consists of a van der Pauw cross for sheet resistance determination merged with a bridge resistor for linewidth measurement. The structure has been analyzed numerically [4] and verified experimentally [5] to yield accurate sheet resistance measurements. Electrical measurement of the linewidth of a conducting layer using the bridge resistor has been described [6]. Each of the cross-bridges included in this test chip was configured to conform to the design rules found elsewhere [7-9] so that systematic geometry-related errors are minimized.

The sheet resistance of the conducting layer is measured by forcing a current I between probe pads A and B while the voltage drop V_R is sensed between probe pads E and F. The sheet resistance, R_S , is determined from

$$R_{s} = f \frac{\pi}{\ln 2} \frac{V_{R}}{I} , \qquad (1)$$

where f is a correction factor, close to unity, that is related to the geometrical asymmetry of the structure. Repetition of the measurement on reversal of the current direction and rotation of the contacts by 90 deg can be made to determine the effects of geometrical asymmetry, measurement voltage offsets, and Joule heating on the measurement [5].

To determine the effective electrical linewidth W_e of a conducting layer, a current I is forced between probe pads A and C and the bridge resistor voltage V_w is sensed between probe pads D and E. The effective width is given by

$$W_{e} = \frac{R_{s}L_{m}}{R^{\star}} = \frac{R_{s}L_{m}I}{V_{w}}, \qquad (2)$$

where R_s is the sheet resistance determined from the top of the cross bridge and L_m is the design length between the bridge voltage taps. Each crossbridge resistor was designed with a bridge length, L_m , of 180 µm.

For the diffused or implanted cross bridges, the effective width W_e is related to the design photomask dimension W_m by

$$W_{e} = W_{m} + \alpha x_{j} \pm W_{e}, \qquad (3)$$

where αx_j is the extent of increase in width due to lateral diffusion and W_e is the amount of oxide over or under etching and the offsets introduced by the photolithographic process [6]. For the metal and polysilicon cross bridges, the lateral diffusion term is omitted in eq (3) and W_e represents the over or under etching of the film and associated photolithographic offsets.

B. Split Bridge Resistor Test Structure

An extension of the cross-bridge resistor, termed the split cross bridge, has been proposed for the measurement of the linewidth, line-to-line spacing, and line-to-line pitch in conducting layers [10]. The structure, seen in figure 4, consists of a polysilicon cross-bridge resistor ($W_b = 20 \ \mu m$, $L_b = 150 \ \mu m$) with the bridge extended and split into two identical conducting channels, W_s ($W_s = 5 \ \mu m$), with a line-to-line spacing of S = 10 \ \mu m. The combined width, W_s' , of the split portion of the bridge is determined by using crossbridge eqs (2) and (3) in section A. The width of the split bridge elements, where $W_s' = 2 \ W_s$, is then subtracted from the width of the unsplit portion W_b to obtain the line spacing S

$$S = W_{b} - 2W_{c}$$
 (4)

The line pitch P is then determined by

$$P = S + W_{S} {.} (5)$$

C. Misregistration Test Structures

The misregistration or misalignment between features on different photomask levels limits the packing density of circuits being manufactured. Errors due to misalignment of the photomask to the wafer, random and systematic alignment errors between the photomasks themselves, and distortions introduced during processing of the photomask and IC all contribute to a total misregistration error that may occur. At least three different misregistration test structures have been developed [11-13], and modified versions of each of three structures have been incorporated in this test chip.

Misregistration between the contact window and gate window photomasks is measured using a test structure similar to the one studied by Russell *et al.* [11] and is shown in figure 5a. The structure is analogous to a pair of linear potentiometers, one oriented vertically and the other horizontally.

The gate window photomask defines the two diffused resistors, joined together by the probe pad F. To measure misregistration, a current I is forced between pad B and pad G and the voltage between the four resistors measured. Perfect alignment of the contact window photomask positions the contact windows for the voltage taps (pads D and J) midway between pads C and E and H and I, respectively. For misregistration between the photomasks, the error in the vertical direction d_y is proportional to the difference between $V_{CD} = V_C - V_D$ and $V_{DE} = V_D - V_E$ and is

$$d_{y} = \frac{W}{2IR_{g}} \left(v_{DE} - v_{CD} \right) = \frac{L}{2} \frac{v_{DE} - v_{CD}}{v_{AC}} .$$
 (6)

Similarly, in the horizontal direction the misregistration d_x is

$$d_{x} = \frac{L}{2} \frac{V_{JI} - V_{HJ}}{V_{AC}}$$
 (7)

The results are usually displayed in the form of a vector map and can be analyzed to determine the sources of the misregistration error [13-15,17].

An analogous structure for measuring misalignment between the metal and final polysilicon photomasks is seen in figure 5b. The contact windows have been enlarged along the length of the resistors in both cases, while the overlapping metal has been narrowed so that the misregistration of the metal photomask rather than the contact window photomask is sensed. The method of measurement is the same as for the previous structure.

Figures 6a, b, and c show three different test structures incorporated in NBS-40, each designed to measure misregistration between the final polysilicon photomask and the gate window photomask. The final structure, seen in figure 6a, is a modification of that seen in figure 5a. The position of the voltage taps (which are connected to pads D and J) are determined by the alignment of the final polysilicon photomask. The same fabrication technique is used to construct the misregistration structure seen in figure 6b following the approach developed by Perloff [12,16,17]. The location of the oversize polysilicon photomask and the final level polysilicon photomask determines the position of the voltage taps. To determine misregistration, a current I is forced between pad E and pad F and the voltage measured between voltage taps C and B and B and D. The misregistration in the horizontal direction is

$$d_{x} = 0.3159 \alpha \arcsin \frac{V_{CB} - V_{BD}}{V_{CB} + V_{BD}}, \qquad (8)$$

where $\alpha = 40 \ \mu m$ is the width of the body of the structure, $V_{CB} = V_C - V_B$, and $V_{BD} = V_B - V_D$. Similarly, misregistration in the vertical direction is

$$d_{y} = 0.3159 \alpha \arcsin \frac{V_{CA} - V_{AE}}{V_{CA} + V_{AE}} .$$
 (9)

The structure also provides for sheet resistance determination from the four voltage differences measured above and the current as given by

$$\rho_{\rm s} = \frac{\pi}{\ln 2} \frac{V_{\rm CB} + V_{\rm BD} + V_{\rm CA} + V_{\rm AE}}{21} \,. \tag{10}$$

A third misregistration test structure designed to measure misregistration between the same two photomask levels as the previous two structures is the Stickman-type structure seen in figure 6c. The structure is a modification of that of Nicholas *et al.* [15,18] in which the current taps and voltage taps are separated in order to provide a four-terminal or Kelvin measurement of resistance. For this design, a long blocking bar on a second photomask is aligned with a larger rectangle on the first photomask level to form two parallel and narrow resistors of equal size for no misregistration. The blocking bar is the final size polysilicon that is aligned with the larger, rectangular gate window opening. The subsequent source-drain implant then forms the two identical resistors. This structure and an identical structure rotated by 90 deg can be used to measure the misregistration between polysilicon and gate window in both the x- and y-directions.

The measurement and analysis is similar to that of the previous misregistration structures. A current I is forced between pad A and pad B while the voltage drops between probe pads F and E and C and D along the bridge resistors are sensed. The displacement in the y-direction is given by

$$d_{y} = \frac{I\rho_{s}L}{2} \left(v_{FE}^{-1} - v_{CD}^{-1} \right) , \qquad (11)$$

where L = 160 μ m, ρ_S is the sheet resistance obtained from an adjacent crossbridge structure, and $V_{FE} = V_F - V_E$, and $V_{CD} = V_C - V_D$. Similarly, the x misregistration component is given by

$$H_{x} = \frac{I\rho_{s}L}{2} \left(V_{GI}^{-1} - V_{JH}^{-1} \right) .$$
 (12)

The availability of the three polysilicon-to-gate-window misregistration structures located immediately adjacent to one another provides the opportunity for a close comparison of the three structures and associated measurement methods.

D. Contact Resistance Test Structures

The six contact resistor test structures [19,20] were incorporated in the NBS-40 test chip to measure interfacial contact resistance $R_C(\Omega)$ for the metal-to-diffusion and metal-to-polysilicon contacts. Three test structures of each type with square contacts of 5, 10, and 20 µm on a side were included in the test chip. A metal-to-diffusion and a metal-to-polysilicon contact resistor test structure with a design $100-\mu m^2$ contact window area can be seen in figures 7a and 7b, respectively. The six-terminal structure [19] allows measurement in the interfacial contact resistance and the "end" contact resistance and the subsequent calculation of the "front" contact resistance

using the transmission line formulation of Berger [21]. The structure and measurement method were designed to minimize the influence of parasitic resistances and contact-to-diffusion misalignment on the result. For a current I forced beween pad A and pad E, the ratio of the voltage difference between pads C and F to the current gives the interfacial contact resistance [22]. Similarly, the ratio of the voltage difference between pads D and F to the current gives the end contact resistance. Uniformity in the interfacial layer can be estimated by using all three structures of the same type, since the interfacial contact resistance is expected to be inversely proportional to the contact window area for a uniform contact layer [22]. The specific contact resistance $\rho_{\rm C}(\Omega \cdot {\rm cm}^2)$, for a uniform interfacial layer, the ratio of the voltage drop across the interface to the current density through it, can be calculated from

$$R_{c} = \rho_{c}/A \quad (13)$$

E. Dopant Profiling Measurements

A large, -channel silicon gate MOSFET, seen in figure 8, was included for profiling the boron threshold adjust implant using a dc current measurement technique. For the device operating in the linear region at a fixed drain-source voltage, the gate-to-source bias, $V_{\rm GS}$, is adjusted to obtain a predetermined current level. Subsequently, as $V_{\rm GS}$ is changed, the back gate bias, $V_{\rm SB}$, is adjusted to maintain a constant drain current. The dopant profile N(W) can be determined by

$$N(W) = \frac{C_{ox}^{2}}{qK_{s}} \left(\frac{d^{2}V_{SB}}{dV_{GS}^{2}}\right)^{-1}$$
(14)

for the depth W

$$W = \frac{K_{\rm S}}{C_{\rm ox}} \left(\frac{dV_{\rm SB}}{dV_{\rm GS}} \right) , \qquad (15)$$

where C_{OX} is the gate oxide capacitance per unit area, K_S is the relative permittivity of silicon, V_{GS} is the gate-to-source voltage, and V_{SB} is the back gate bias. In general, profiles determined by this technique are limited to distances from the surface greater than three times the Debye length for the material [23]. Further description of the measurement technique and analysis is found elsewhere [23-27]. The design dimensions for the MOSFET gate are L = 10 µm and W = 440 µm. A polysilicon gate capacitor, see figure 9a, has been included on the chip to permit experimental determination of the oxide capacitance per unit area C_{OX} (area = 2.94 × 10⁵ µm²) of the gate oxide region. The metal gate capacitor, see figure 9b, (area = 7.35 × 10⁵ µm²) can be used to determine the field oxide thickness.

The boron enhancement implant can be profiled from the polysilicon gate MOS capacitor using the capacitance-voltage measurement [28,29]. For lightly

doped substrates, the effect of series resistance on the measurement should be considered [30].

For the C-V technique, the dopant concentration N(W) can be determined by

$$N(W) = \left[qK_{s} \frac{1}{C_{m}} \frac{d}{dV_{GS}} \left(\frac{1}{C_{m}}\right)\right]^{-1} , \qquad (16)$$

for a depth

$$W = K_{s} \left[\frac{1}{C_{m}} - \frac{1}{C_{ox}} \right] ,$$
 (17)

where C_m is the capacitance measured in depletion at a gate bias V_{GS} . The limitations of the C-V method are described in detail by Nicollian and Brews [29].

F. Gated Diode Test Structures

Both a gated diode [31,32] and a gated diode electrometer [33,34], seen in figures 10a and 10b, respectively, have been included in the test chip for measurement of generation lifetime and surface leakage current. In the former, direct measurements of the low, reverse leakage current of the p-n junction under various biases on the MOS gate are used to separate the bulk, the surface depletion region, and the interface generation components. For the first component, an associated bulk generation lifetime $\tau_{\rm G}$ can be derived using

$$\tau_{\rm G} = \frac{qn_{\rm i}^{\rm W} A_{\rm MJ}}{2I_{\rm G}} , \qquad (18)$$

where q is the electronic charge, n_i is the intrinsic carrier concentration, I_G is the reverse leakage current of the junction for the surface region in accumulation, W is the junction depletion width, and A_{MJ} is the junction surface area ($A_{MJ} = 5.52 \times 10^4 \mu^2$).

When the surface under the MOS gate is inverted, the reverse leakage current consists of two components: I_G , and the current I_{GI} due to the induced depletion region under the gate. The generation lifetime in the vicinity of the surface, τ_S , can then be obtained from

$$F_{s} = \frac{qn_{i}W_{I}A_{I}}{2I_{GI}}, \qquad (19)$$

where A_I is the area under the gate ($A_I = 8.21 \times 10^3 \ \mu m^2$) and W_I is the depletion width of the field-induced depletion region.

A gated diode electrometer is included in the test chip. The gated diode previously described is incorporated into this structure. The electrometer is intended to reduce off-chip loading effects encountered in gated diode measurements. A schematic circuit diagram for the electrometer is seen in figure 10c where a MOSFET operating in the saturated mode is used as the load resistor $R_{\rm L}$.

To analyze the measured dV_O/dt output of the electrometer, the p-n junction capacitance C, gate-source capacitance C_{GS} , load resistance R_L , and transconductance g_m of the source follower transistor are needed. The junction capacitance C can be estimated using the known junction area given above and substrate doping or measured directly on the gated diode. The gate-source capacitance can only be estimated from the known geometry of the self-aligned MOSFET (W = L = 20 µm), oxide thickness, and gate-source overlap nominally the source-drain implant junction depth. The transconductance g_m of the enhancement MOSFET can be obtained directly from measurement on an identical MOSFET (W = L = 20 µm) in column 1 of the chip. The saturated MOSFET used as the load resistor R_L can be measured directly between pads C and E of the electrometer to obtain R_L .

G. Capacitance Measurements

The polysilicon gate capacitor (see fig. 9a) can be used to determine the gate oxide capacitance per unit area, gate oxide thickness, and flat band capacitance and voltage from C-V measurements. It can also be used to determine Debye length [35], mobile ion density [36], oxide fixed charge [37], and interface trap density [38]. The generation lifetime can be measured using a transient capacitance technique [39] or a modified linear sweep technique [40,41].

H. Mobility Measurements

An enlarged (W = 240 μ m, L = 35 μ m) silicon gate MOSFET seen in figure 11 was included to measure the electron inversion layer mobility in the *n*-channel enhancement MOSFETs. The effective electron mobility is obtained from the dc drain conductance g_d measured in the linear region [42] using

$$\mu_{\text{eff}} = \frac{(L/W) g_{d}}{q N_{\text{inv}}}, \qquad (20)$$

where W and L are the design MOSFET channel width and length, and q N_{inv} is the total induced charge in the channel per unit area. This relation is only valid for long-channel devices. By employing a large MOSFET, errors in determining actual device geometries are minimized. The determination of μ_{eff} using this approach does require that the gate capacitance per unit area be obtained from the silicon gate capacitor. The degradation of the mobility with increasing gate voltage [42,43] can be quantified using this measurement.

I. MOSFET Test Structures

A metal gate thick oxide MOSFET and a polysilicon gate thick oxide MOSFET, seen in figure 12, with $W = L = 30 \ \mu m$ are included in the test chip for field threshold voltage determination. The silicon gate MOSFET will provide the threshold voltage for the polysilicon over the thermally grown field oxide; the metal gate will provide the threshold voltage of the metal over the total field oxide which consists of the thermally grown field oxide and the oxide deposited over the polysilicon.

Two sets of silicon gate MOSFETs seen in figures 13a and b are included. One set consists of enhancement devices with fixed width (W = 20 μ m) and differing lengths (L = 2.0 μ m, 5.0 μ m, 10 μ m, and 20 μ m). The other set consisting of depletion devices has a fixed length (L = 50 μ m) and varying widths (W = 2.0 μ m, 5.0 μ m, 10 μ m, and 20 μ m). The set of enhancement devices can be used to electrically estimate the effective channel length [42,43] and to determine the total source, drain, and contact resistance [45].

The measurement of MOSFET parameters such as threshold voltage, transconductance, drain resistance, body effect, channel length modulation, breakdown, interface trapped charge, etc., are described in a number of publications [42,48-52].

J. Circuit Elements

An inverter seen in figure 14 consisting of an enhancement driver with $W = 20 \mu m$, L = 10 μm and a depletion load with W = 10 μm , L = 50 μm is included. Measurement of the static transfer characteristic of the inverter can be used to obtain the logic level voltages and noise margins. A ring oscillator consisting of 19 inverter stages with a NAND startup and a four-stage, separately powered output buffer [53] is seen in figure 15. Both the propagation delay and power dissipation per stage can be determined. Also, the ring oscillator can be used to evaluate the sensitivity of circuit codes to process-induced variations in input parameters [54]. The initial stage of the output buffer is the same inverter used in the ring. Each subsequent stage has the channel width of both the driver and load enlarged by a factor of three over the previous stage so that load capacitance of 20 pF can be driven without an excessive propagation time penalty [55].

4. SUMMARY

An NMOS test chip has been described that includes 35 test structures for the measurement of a variety of device, fabrication process, and circuit parameters associated with the fabrication of ICs. In general, the test structures and measurement methods described assume long-channel device characteristics. Structure layout, the method of parameter measurement, and references to the literature are provided for each structure.

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TABLE

Test Chip NBS-40

Device #	Test Structure	Parameters Determined
1.	Enhancement MOSFET array with $W = 20 \ \mu m$, $L = 2.0, 5.0, 10, 20 \ \mu m$	threshold voltage, transcon- ductance, effective channel length and width
2.	Depletion MOSFET array with L = 50 μ m, W = 2.0, 5.0, 10, 20 μ m	
3.	Polysilicon cross bridge, W = 10 μ m, L = 180 μ m	polysilicon sheet resistance, linewidth
4.	Polysilicon cross bridge, $W = 5 \mu m$, L = 180 μm	polysilicon sheet resistance, linewidth
5.	Metal cross bridge, W = 10 μm, L = 180 μm	metal sheet resistance, linewidth
6.	Contact window to gate window misregistration (NBS) L = 160 um	misregistration between contact window and gate window
7.	MOSFET L = 10 μ m, W = 440 μ m	dopant profile
8.	Polysilicon-to-gate window misregistration (Perloff) $\alpha = 40 \ \mu m$	misregistration between final polysilicon and gate window
9.	Polysilicon-to-gate window misregistration (NBS) L = 160 µm	misregistration between final polysilicon and gate window
10.	Split cross-bridge resistor	line-to-line spacing between conductors, line-to-line pitch
11.	Polysilicon-to-gate window misregistration (Stickman) L = 160 µm	misregistration between final polysilicon and gate window
12.	n^+ diffused cross bridge, W = 10 µm, L = 180 µm	n^+ sheet resistance, linewidth
13.	n^+ diffused cross bridge, W = 5 µm, L = 180 µm	n^+ sheet resistance, linewidth

Device #	Test Structure	Parameters Determined
14.	Metal-to-polysilicon misregistration (NBS) L = 160 µm	misregistration between metal and and polysilicon
15.	Metal-polysilicon contact resistance, 20-µm square contact	interfacial, front, and end contact resistance, interfacial layer uniformity
16.	Metal-polysilicon contact resistance, 10-µm square contact	
17.	Metal-polysilicon contact resistance, 5-µm square contact	"
18.	Metal and polysilicon gate field oxide MOSFETs with W = L = 30 µm	field threshold voltage
19.	Gated diode $A_{MJ} = 5.52 \times 10^4 \ \mu m^2$ $A_I = 8.21 \times 10^3 \ \mu m^2$	generation lifetime, surface leakage current
20.	Metal- n^+ contact resistance, 20-µm square contact	interfacial, front, and end contact resistance, interfacial layer uniformity
21.	Metal- n^+ contact resistance, 10-µm square contact	"
22.	Metal- n^+ contact resistance, 5-µm square contact	и
23.	Gated diode electrometer	generation lifetime, surface leakage current
24.	Enhancement MOSFET $W = 240 \ \mu m$, L = 35 μm	inversion layer mobility
25.	Polysilicon gate MOS capacitor A = $2.94 \times 10^5 \ \mu m^2$	oxide thickness, dopant profile, oxide charge
26.	Seventeen-stage ring oscillator with output buffer	propagation delay, power dissipation
27.	Single-stage inverter of ring oscil- lator, driver W = 20 µm, L = 10 µm load W = 10 µm, L = 50 µm	transfer characteristic, voltage levels, noise margins
28.	Metal gate capacitor A = $7.35 \times 10^5 \ \mu m^2$	oxide thickness, dopant profile, oxide charge





Figure 2. Cross section of an *n*-channel silicon gate enhancement MOSFET.

Photomask Level

Description

1

2

3

4

5

6

7

8

 p^+ isolation diffusion

gate window

enhancement threshold adjust implant (boron)

depletion threshold adjust implant (phosphorus)

oversize polysilicon

final size polysilicon

contact window

aluminum metallization

9

passivation













Figure 3a. Polysilicon cross-bridge resistor.



Figure 3b. Diffused cross-bridge resistor.







Figure 4. Polysilicon split-bridge resistor.















Figure 6b. Polysilicon-to-gate-window misregistration test structure (Perloff).



Figure 6c. Polysilicon-to-gate-window misregistration test structure (Stickman).



Figure 7a. Six-terminal metal-to-diffusion contact resistor.



Figure 7b. Six-terminal metal-to-polysilicon contact resistor.



Figure 8. *n*-channel enhancement MOSFET for dopant profile determination, L = 10 μ m, W = 440 μ m.



Figure 9a. Polysilicon gate capacitor, A = $3.06 \times 10^5 \ \mu m^2$.







Figure 10a. Gated diode.



Figure 10b. Gated diode electrometer.



Figure 10c. Schematic drawing of the gated diode electrometer where GD is the gated diode, Q1 the reset switch, Q2 the source-follower, and R_L is the load resistor.



Figure 11. *n*-channel enhancement MOSFET, $L = 35 \ \mu m$, $W = 240 \ \mu m$.



Figure 12. Metal-gate, field oxide MOSFET (top) and polysilicon-gate field oxide MOSFET (bottom).









Figure 13a. *n*-channel enhancement MOSFET array, L = 2.0, 5.0, 10.0, and 20.0 μ m, W = 20 μ m.









Figure 13b. *n*-channel depletion MOSFET array, L = 2.0, 5.0, 10.0, and 20.0 μ m, $W = 50 \ \mu$ m.







Figure 15. Ring oscillator.

APPENDIX NMOS Process for NBS Test Chip

Wafer: $\rho = 20 \ \Omega \cdot cm$, *p*-type (100) Diffusion tube inside diameter: 75 mm 1. Initial wafer clean 2. Initial oxidation Furnace temperature: 1100°C O₂ flow rate: 500 cc/min Cycle: 20 min dry 20 min wet $t_{ox} = 0.33 \, \mu m$ 20 min dry 3. Isolation lithography (photomask #1) 4. Boron predeposition Furnace temperature: 975°C N₂ flow rate: 500 cc/min Cycle: 15 min 5. Boron skin removal 700°C Furnace temperature: O₂ flow rate: 300 cc/min 20 min Cycle: 6. Boron drive-in and reoxidation 1100°C Furnace temperature: 500 cc/min O₂ flow rate: Cycle: 20 min dry 40 min dry 120 min dry 7. Gate window lithography (photomask #2) 8. Gate oxide growth (75 nm) Furnace temperature: 1000°C O₂ flow rate: 500 cc/min N_2^- + TCE flow rate: 25 cc/min Cycle: 110 min dry 10 min N₂ anneal $t_{ox} = 0.076 \, \mu m$ 9. Enhancement threshold implant lithography (photomask #3)

10. Enhancement threshold implant (target $V_{TE} = 1.0$ V)

Ion:	Boron
Ion energy:	60 keV
lose:	$5 \times 10^{11} \text{ ions/cm}^2$

11. Depletion threshold implant lithography (photomask #4)

12. Depletion threshold implant (target $V_{TD} = -3.0$ V)

Ion:	Phosphorus
Ion energy:	120 keV
Dose:	$6 \times 10^{11} \text{ ions/cm}^2$

13. Polysilicon deposition

Temperature:	650°C	
Time:	13 min	
Thickness:	0.8 µm	

- 14. Oversize polysilicon lithography (photomask #5)
- 15. Phosphorus predeposition for source-drain and polysilicon gate doping

Fur	nace	temperature:	9509	°C
N ₂	flow	rate:	500	cc/min
0_2	flow	rate:	50	cc/min
Cyc	:le:		14	min

16. Phosphorus drive-in and reoxidation

Furnace	temperature:	10009	°C	
0_2 flow	rate:	500	cc/n	nin
Cycle:		18	min	dry

- 17. Polysilicon gate lithography (photomask #6)
- 18. Phosphorus source-drain implant

Ion energy:	120 keV	
Dose:	2×10^{13} ion:	s/cm ²

19. Implant activation

Furnace	temperature:	900	°C
N ₂ flow	rate:	500	cc/min
Cycle:		20	min

20. Oxide deposition

Temperature:	800°C
Time:	30 min

- 22. Al deposition
- 23. Al interconnect lithography (photomask #8)
- 24. A2 alloying

Furnace	temperature:	5009	°C
N, flow	rate:	250	cc/min
Cycle:		3	min

- 25. P SiO₂ glass deposition
- 26. Passivation lithography (photomask #9)

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